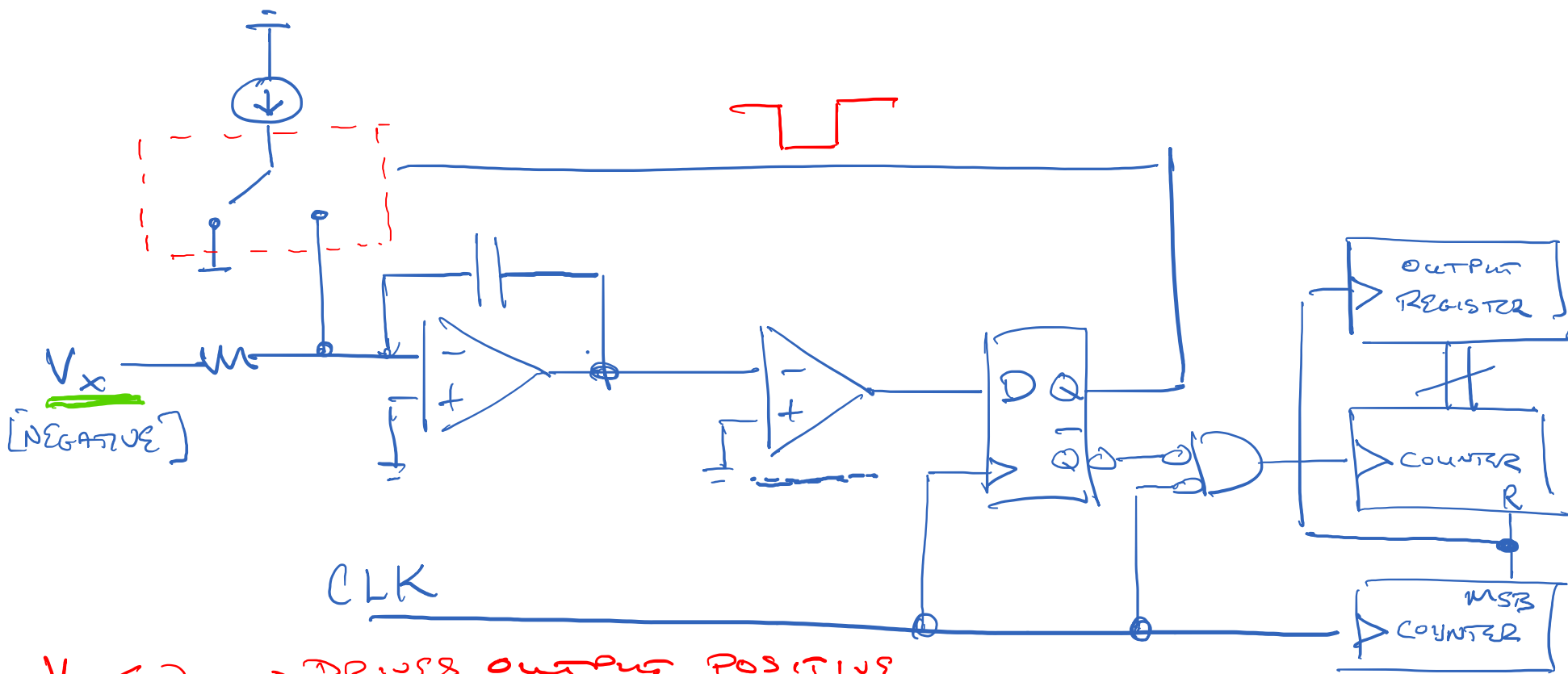
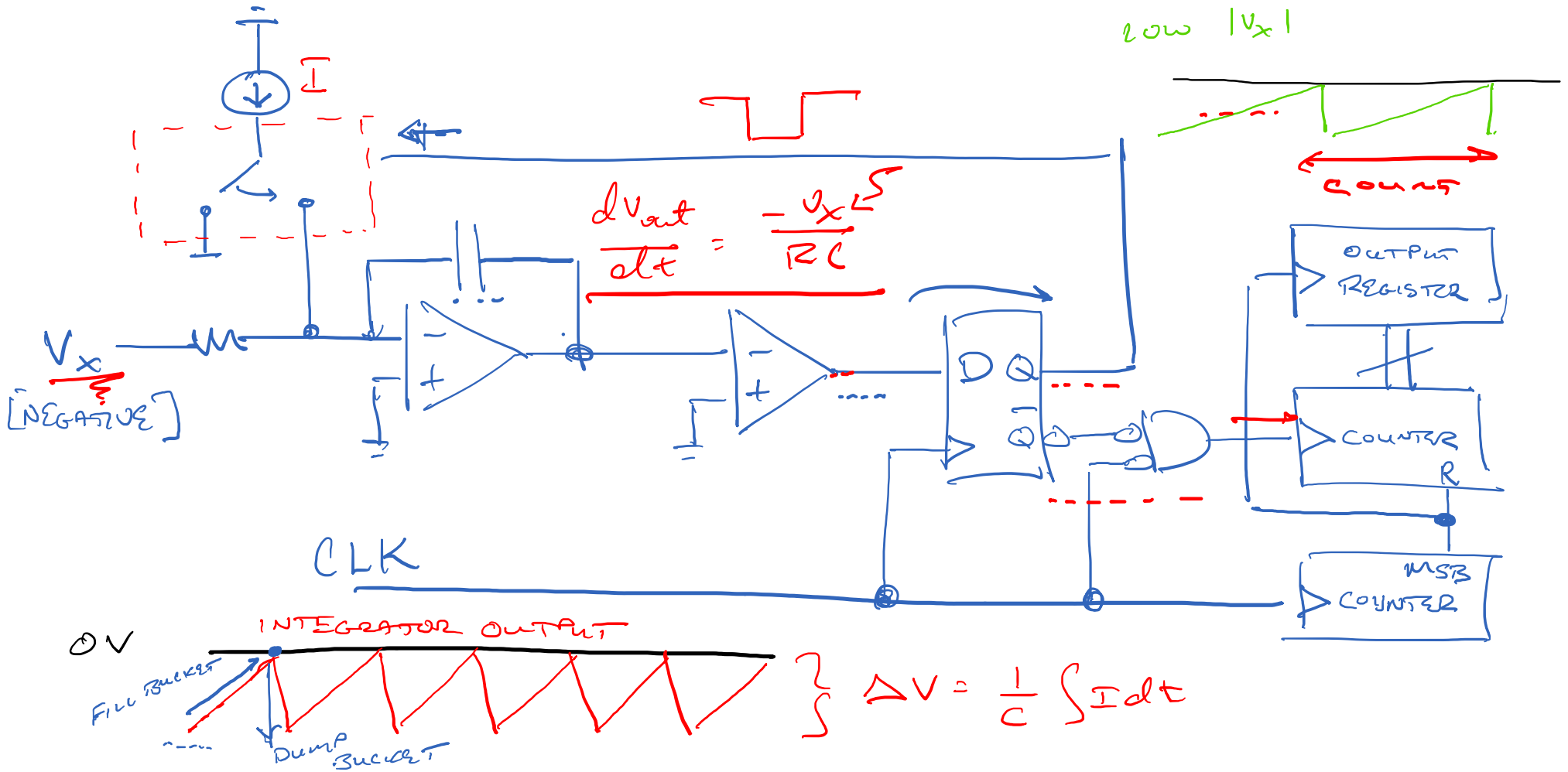


CHARGE-BALANCING ADC Δ - Σ

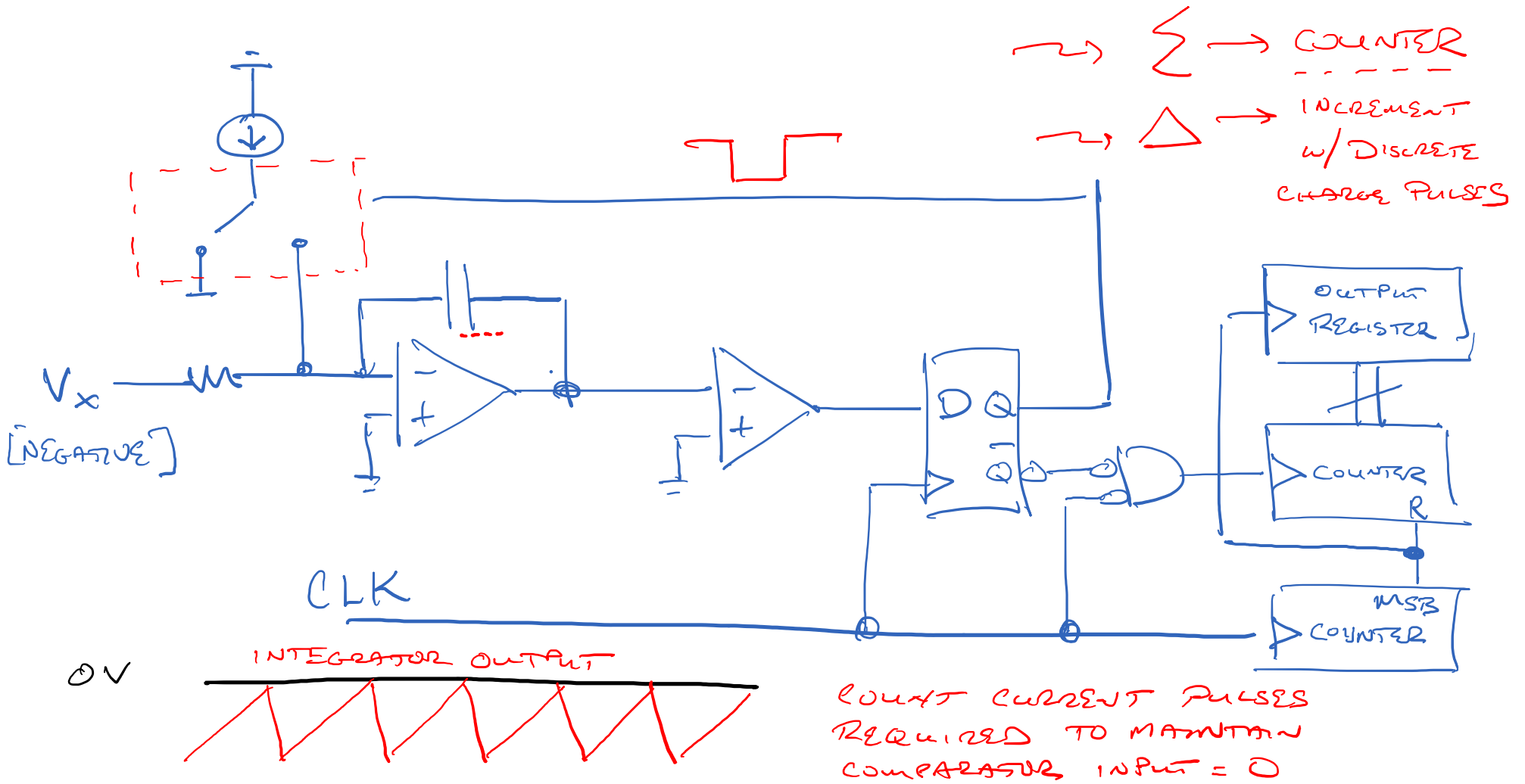


$V_x < 0 \rightarrow$ DRIVES OUTPUT POSITIVE
 DRIVES COMPARATOR NEGATIVE
 THIS CAUSES $Q \rightarrow 0$ [INJECT CHARGE BALANCING PULSE]

CHARGE-BALANCING ADC Δ - Σ

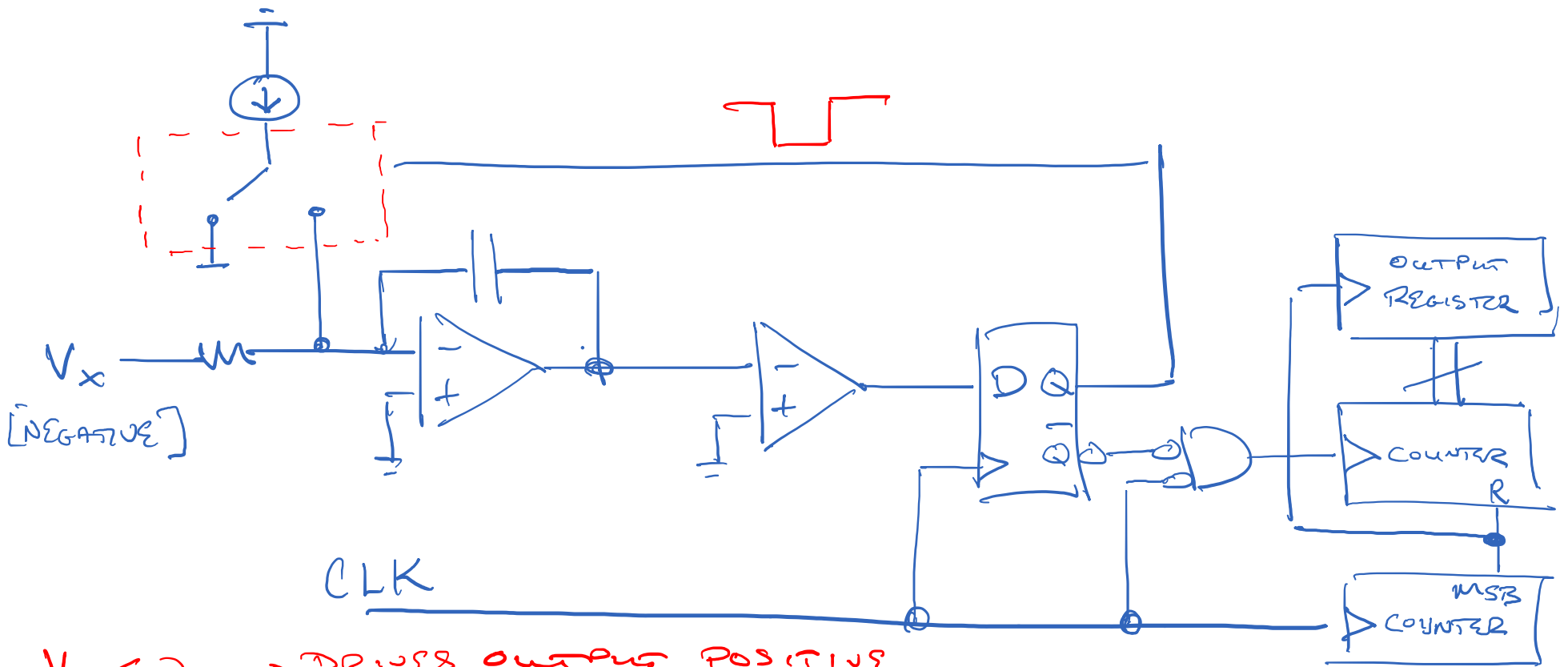


CHARGE-BALANCING ADC Δ - Σ



ROUND CURRENT PULSES
REQUIRED TO MAINTAIN
COMPARATOR INPUT = 0

CHARGE-BALANCING ADC Δ - Σ



$V_x < 0 \rightarrow$ DRIVES OUTPUT POSITIVE
 DRIVES COMPARATOR NEGATIVE
 THIS CAUSES $Q \rightarrow 0$ [INJECT CHARGE BALANCING PULSE]

SUMMARY



→ FLASH

→ SINGLE/DUAL SLOPE

{ SAR
ΔΣ

SPEED

FAST [~20ns]

SLOW [~ms]

MEDIUM [~μs]

MEDIUM [1-10μs]

RESOLUTION

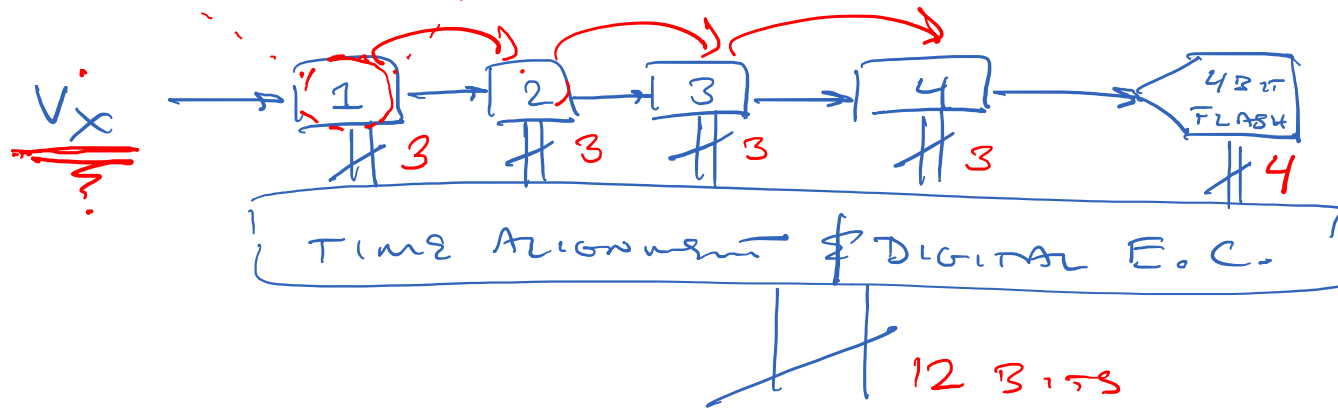
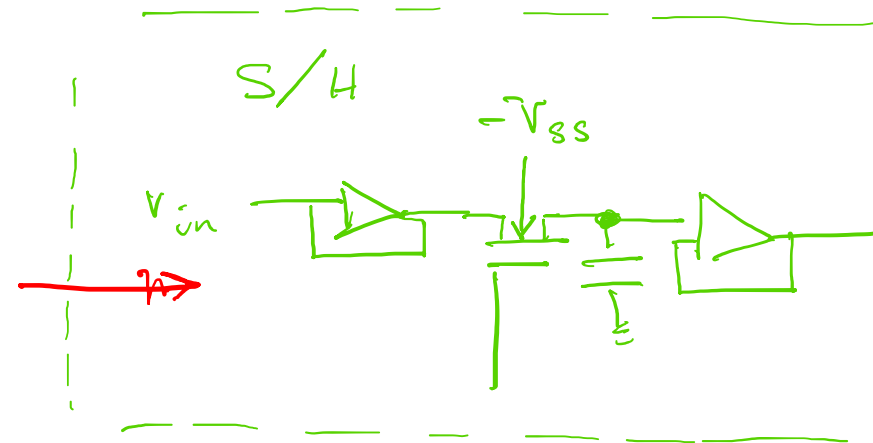
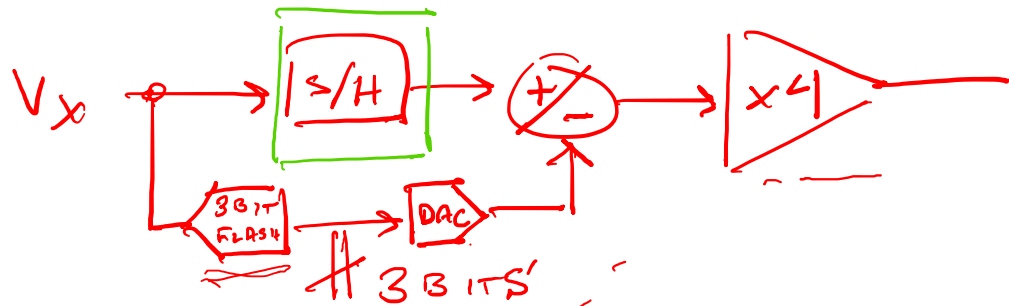
POOR [→ ~10 BIT]

EXCELLENT [~24 BIT]

MODERATE [~16-20 BITS]

MODERATE [~16-20 BITS]

PIPELINED ADC



SEE, eg., maximintegrated.com

CIRCUIT SIMULATION

SPICE [BERKELEY EECS', 1970's]

SIMULATION PROGRAM WITH INTEGRATED CIRCUIT EMPHASIS

DC ANALYSIS

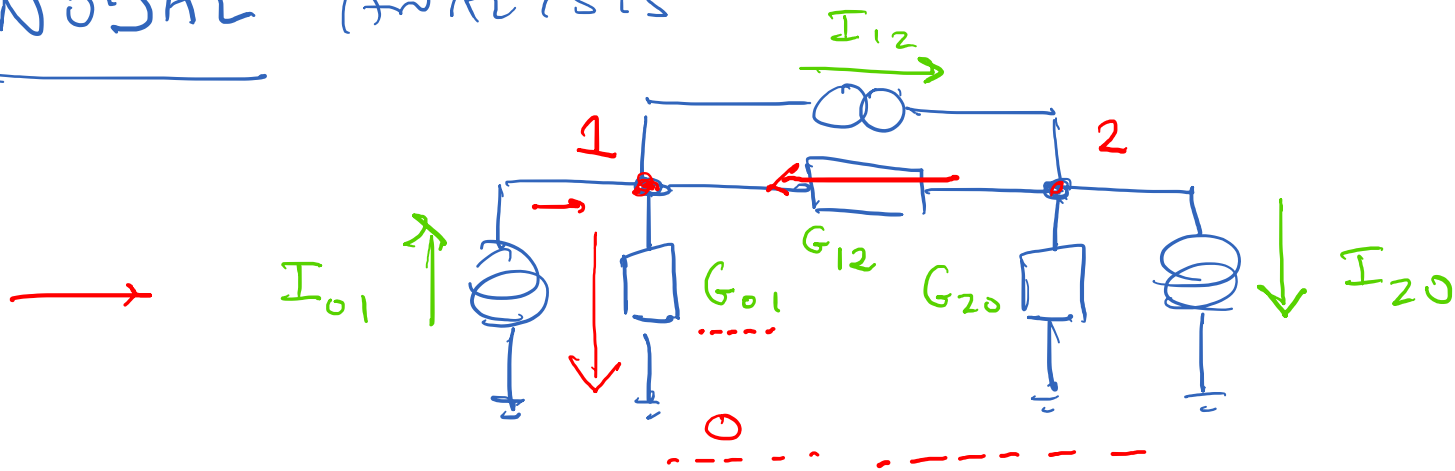
AC ANALYSIS

→ TRANSIENT ANALYSIS

MODELING OF MULTITERMINAL,

NONLINEAR DEVICES, ETC.

NODAL ANALYSIS



$$\sum I = 0$$

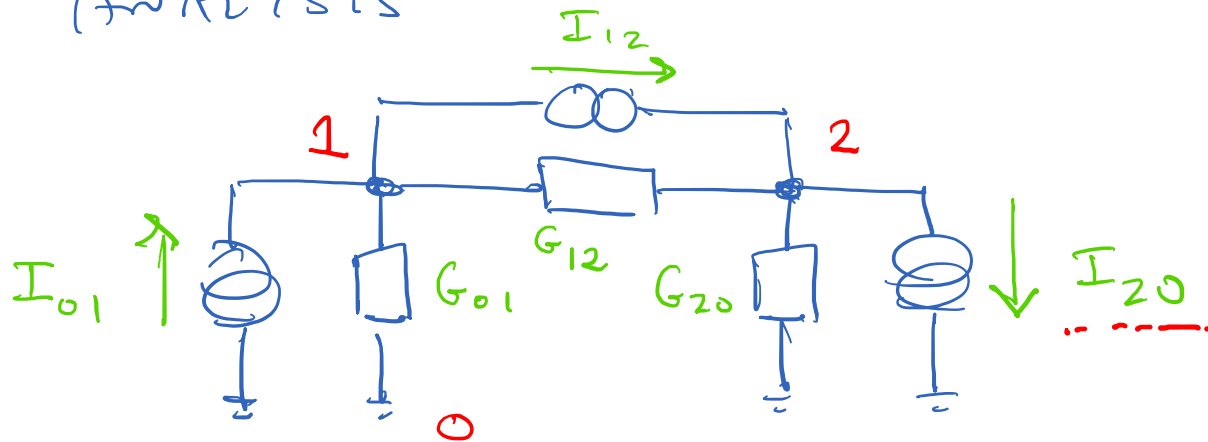
KIRCHHOFF CURRENT LAW:

$$\text{NODE 0} \quad G_{01} (V_1 - V_0) - G_{20} (V_0 - V_2) + I_{20} - I_{01} = 0$$

$$\text{NODE 1} \rightarrow G_{12} (V_2 - V_1) - G_{01} (V_1 - V_0) + I_{01} - I_{12} = 0$$

$$\text{NODE 2} \quad G_{20} (V_0 - V_2) - G_{12} (V_2 - V_1) + I_{12} - I_{20} = 0$$

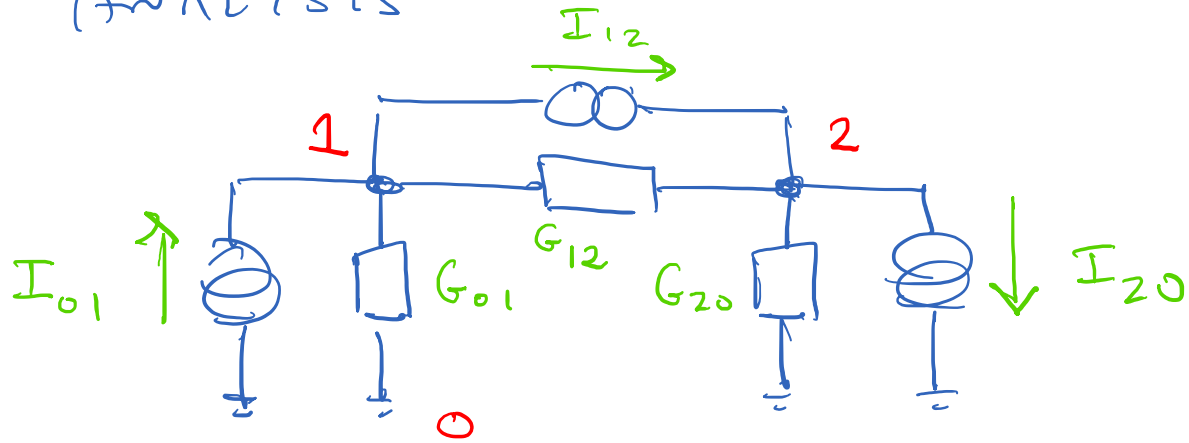
NODAL ANALYSIS



$$\begin{bmatrix} G_{01} + G_{20} & -G_{01} & -G_{20} \\ -G_{01} & G_{12} + G_{01} & -G_{12} \\ -G_{20} & -G_{12} & G_{20} + G_{12} \end{bmatrix} \begin{bmatrix} V_0 \\ V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} I_{20} - I_{01} \\ I_{01} - I_{12} \\ I_{12} - I_{20} \end{bmatrix}$$

$\underline{G} \cdot \underline{V} = \underline{I}$

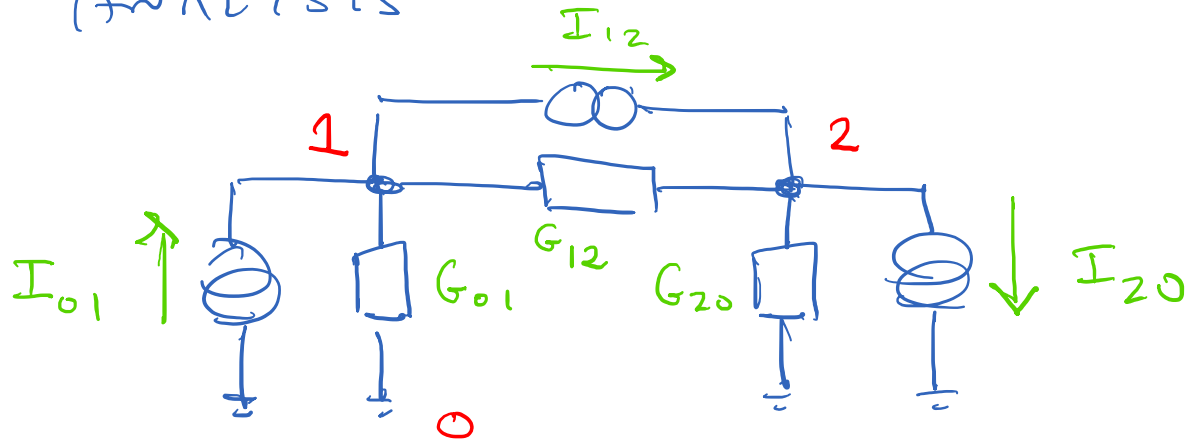
NODAL ANALYSIS



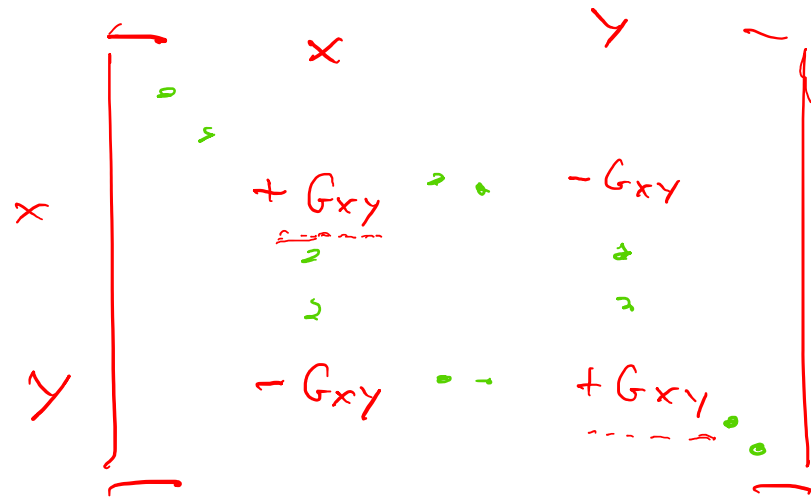
$$\begin{matrix} \rightarrow & \begin{matrix} \parallel \\ G \\ \parallel \end{matrix} & \begin{matrix} \parallel \\ V \\ \parallel \end{matrix} & = & \begin{matrix} \parallel \\ I \\ \parallel \end{matrix} \\ \rightarrow & \begin{matrix} \parallel \\ V \\ \parallel \end{matrix} & = & \begin{matrix} \parallel \\ G \\ \parallel \end{matrix} & - & \begin{matrix} \parallel \\ I \\ \parallel \end{matrix} \end{matrix}$$

\leftarrow Matrix

NODAL ANALYSIS

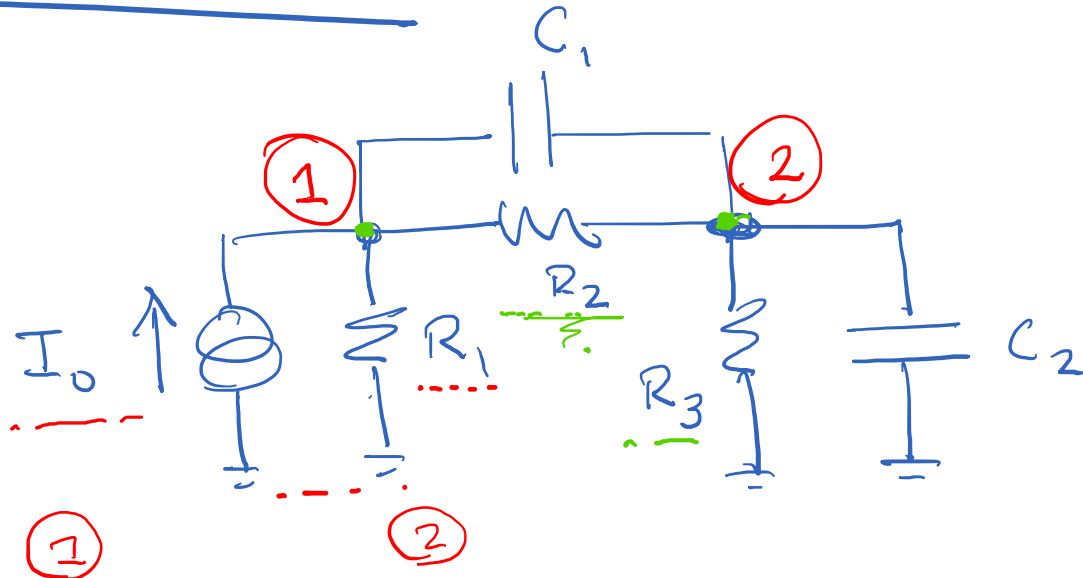


"PATTERN OF FOUR"



CONCRETE EXAMPLE

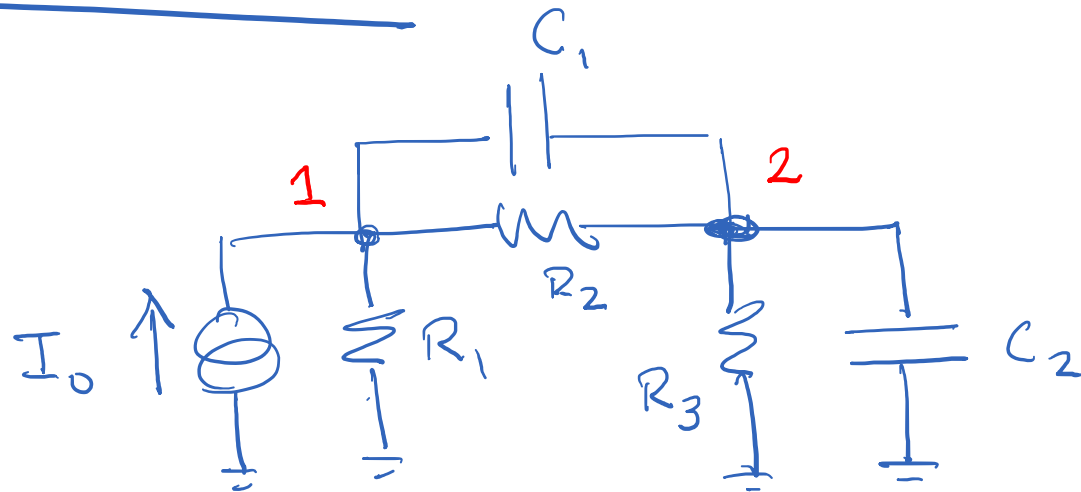
DC ANALYSIS ONLY



$$\begin{matrix} \textcircled{1} \\ \textcircled{2} \end{matrix} \begin{bmatrix} \frac{1}{R_1} + \frac{1}{R_2} & -\frac{1}{R_2} \\ -\frac{1}{R_2} & \frac{1}{R_2} + \frac{1}{R_3} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} I_0 \\ 0 \end{bmatrix}$$

KNOWN KNOWN

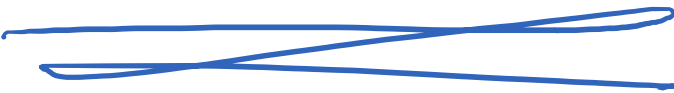
CONCRETE EXAMPLE



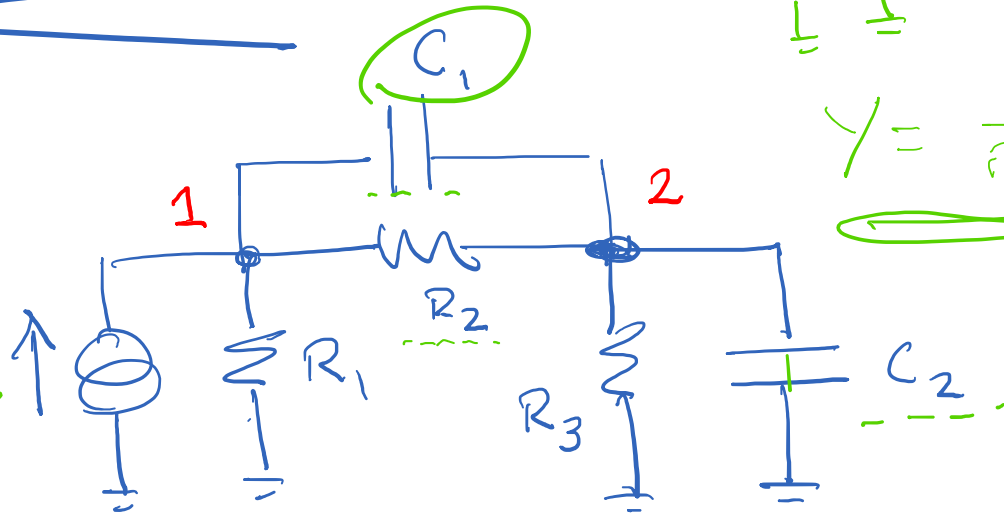
INVERT :

$$\underline{\underline{V}} = \underline{\underline{G}}^{-1} \underline{\underline{I}}$$

EXTENSION TO AC CIRCUITS



$I_0 e^{j(\omega t + \phi)}$



$$\underline{V} = \underline{C}^{-1} \underline{Q}$$

$$Y = \frac{1}{R} + j\omega C$$



$$\begin{bmatrix} Y & V \\ V & = & F \\ \underline{V} = \underline{Y}^{-1} & \underline{F} \end{bmatrix}$$

$$\begin{bmatrix} \frac{1}{R_1} + \frac{1}{R_2} + j\omega C_1 & -\frac{1}{R_2} - j\omega C_1 \\ -\frac{1}{R_2} - j\omega C_1 & \frac{1}{R_3} + \frac{1}{R_2} + j\omega(C_1 + C_2) \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} I_0 e^{j\phi} \\ 0 \end{bmatrix}$$

NONLINEAR CIRCUITS

→ DC ANALYSIS :

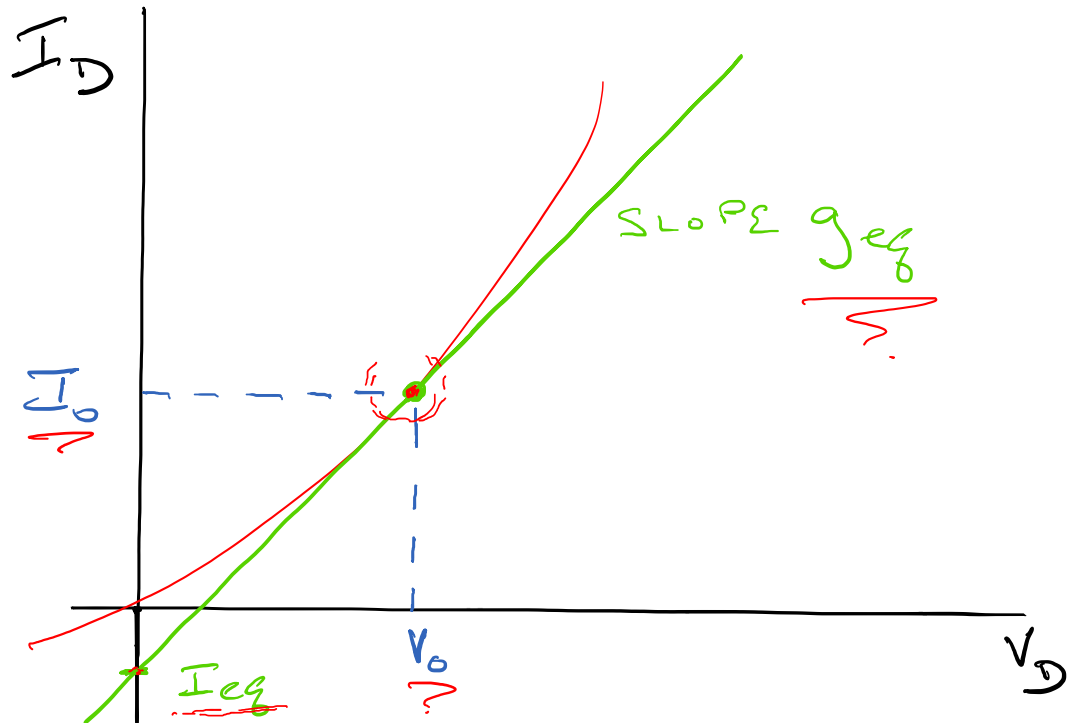
DC SOLUTION OBTAINED BY
ITERATIVE PROCEDURE

DIODE

$$I_D = I_s \left[e^{V_D/V_T} - 1 \right]$$

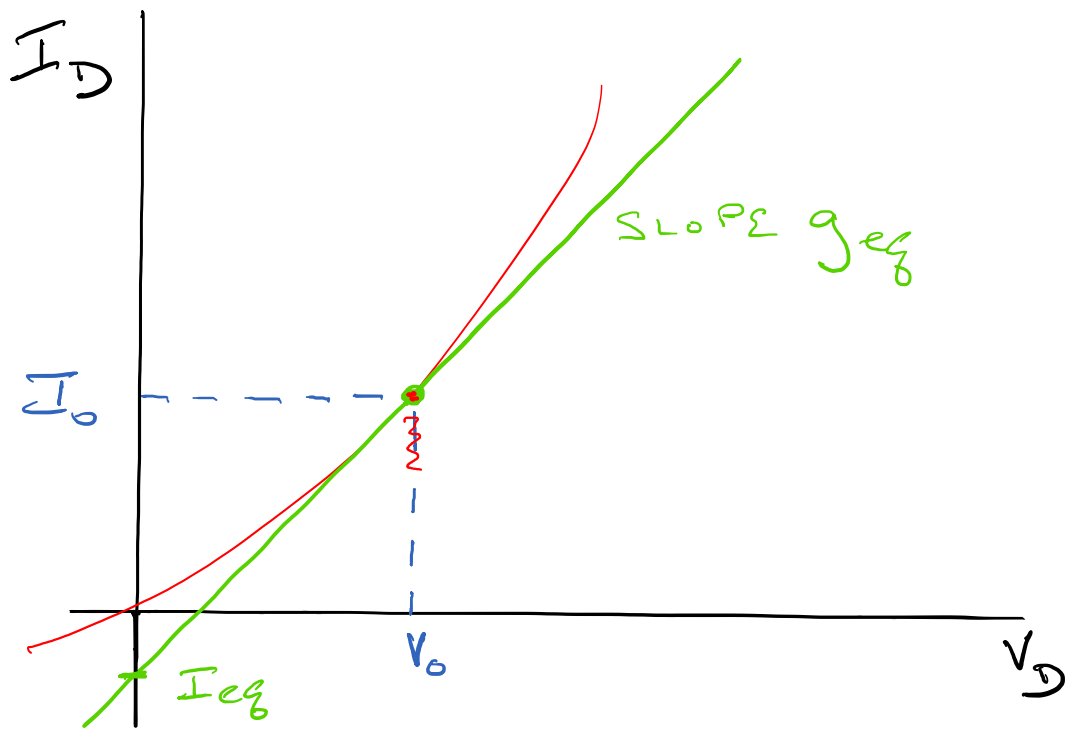
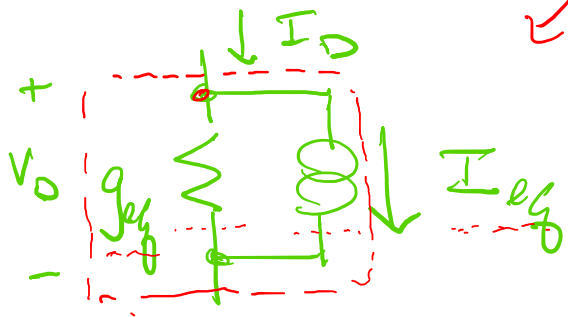
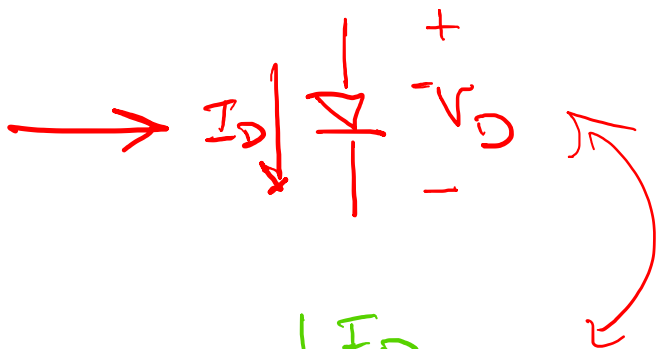
$$g_{eq} = \frac{I_s}{V_T} e^{V_D/V_T}$$

$$I_{eq} = I_0 - g_{eq} V_0$$

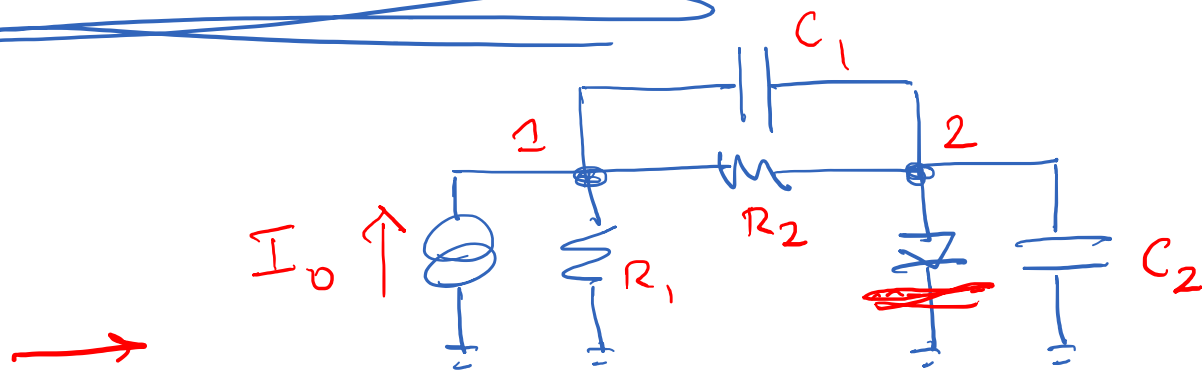


$$g_{eg} = \frac{I_s}{V_T} e^{V_D/V_T}$$

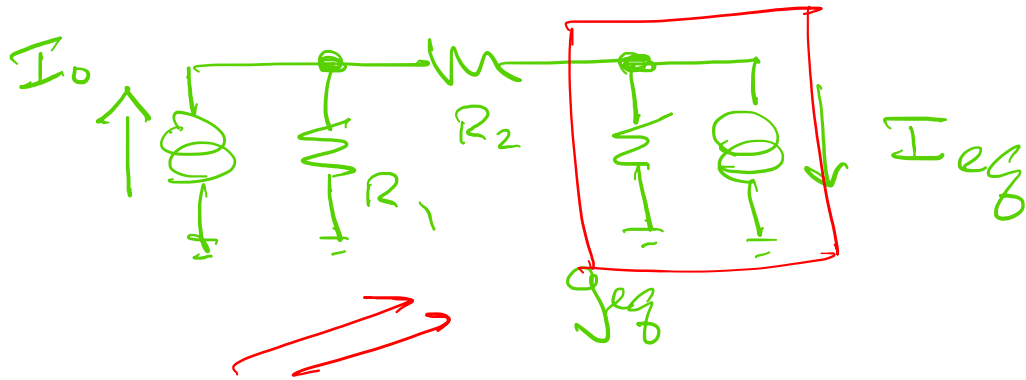
$$I_{eg} = I_0 - g_{eg} V_D$$



CONCRETE EXAMPLE

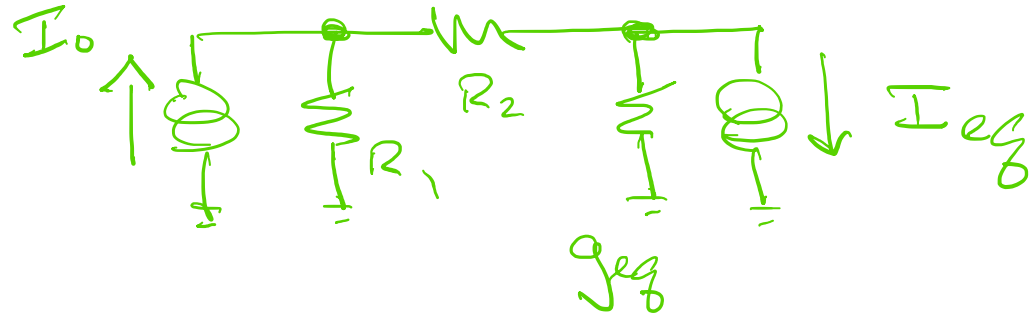


DC
LINEARIZED
EQUIPMENT



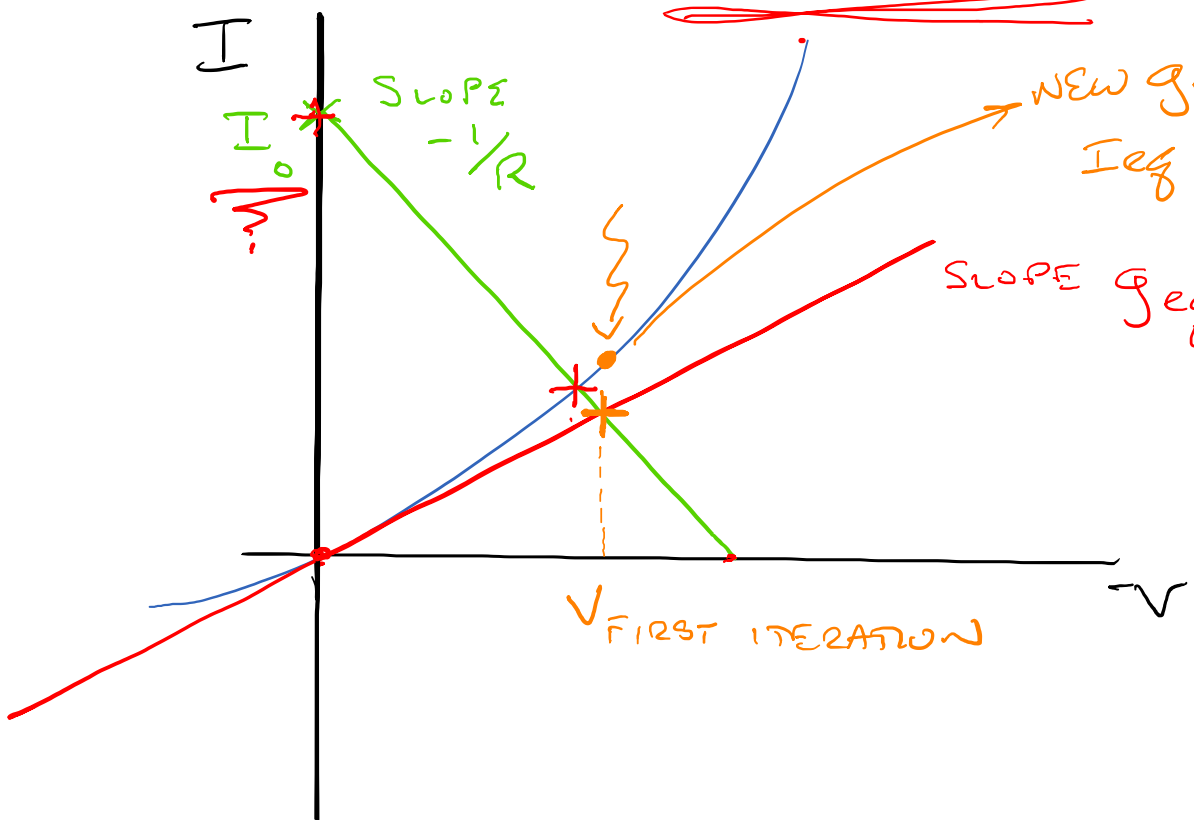
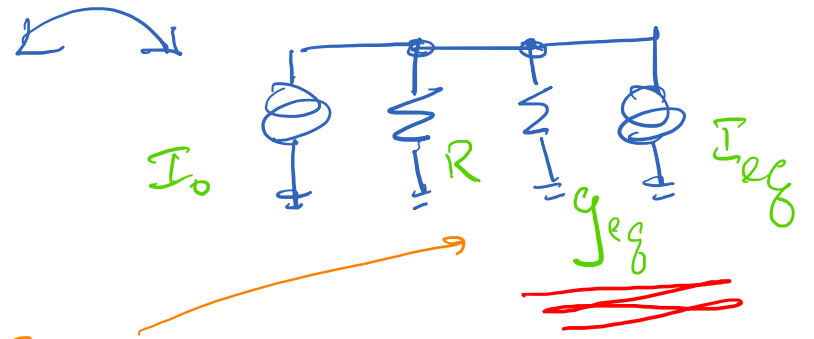
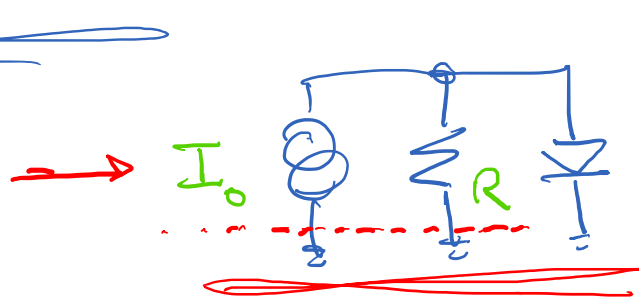
CONCRETE EXAMPLE

DC
LINEARIZED
EQUIPMENT

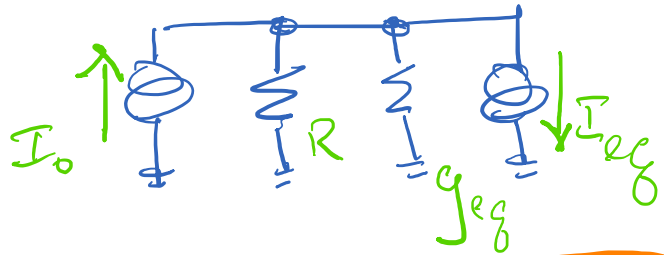


$$\begin{bmatrix} \frac{1}{R_1} + \frac{1}{R_2} & -\frac{1}{R_2} \\ -\frac{1}{R_2} & \frac{1}{R_2} + g_{eq} \end{bmatrix} \begin{bmatrix} V_1 \\ V_0 \end{bmatrix} = \begin{bmatrix} I_0 \\ -I_{eq} \end{bmatrix}$$

EVEN SIMPLER.



FIRST ITERATION:
 $g_{eq}(V=0)$
 $= \frac{I_s}{V} \exp(V)$
 $I_{eq} = 0$



DC ANALYSIS :

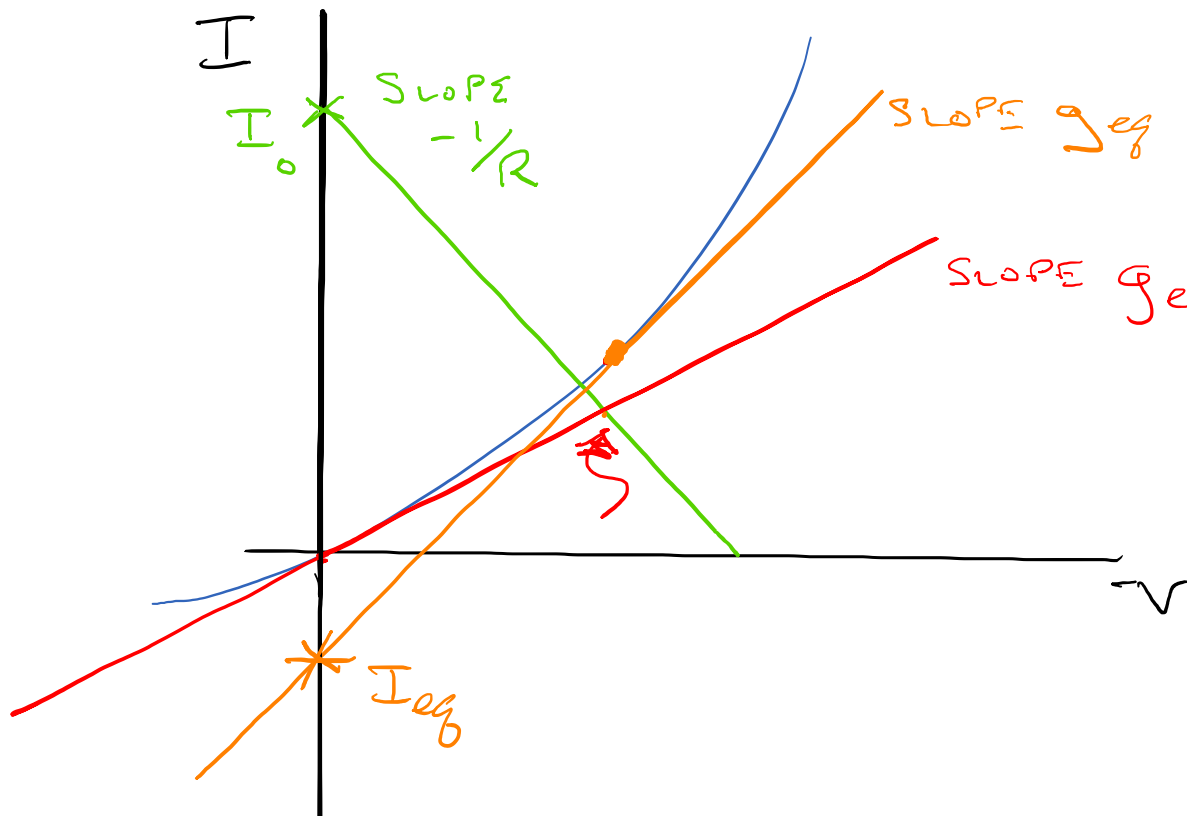
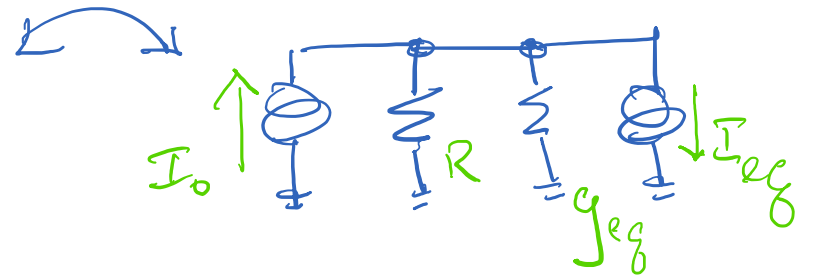
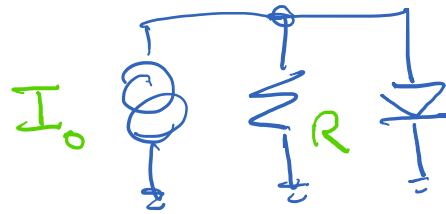
$$\left[\frac{1}{R} + g_{eq} \right] V = \underline{I_0} - \underline{I_{eq}}$$

The term $\left[\frac{1}{R} + g_{eq} \right]$ is underlined with an orange bracket. The terms I_0 and I_{eq} in the right-hand side are also underlined with orange lines. A curly brace is drawn above the equation.

$$V = \frac{I_0 - I_{eq}}{\frac{1}{R} + g_{eq}} \Rightarrow \text{ITERATE}$$

The word "ITERATE" is written in orange and has a curved arrow underneath it pointing back to the equation, indicating an iterative process.

EVEN SIMPLER.



UPDATE g_{eq}, I_{eq}

$$V = \frac{I_0 - I_{eq}}{\frac{1}{R} + g_{eq}}$$

ETC.

ETC.

TRANSIENT ANALYSIS



DISCRETIZE TIME INTERVAL $(0, T) \rightarrow (0, t_1, t_2, \dots, T)$

NUMERICAL INTEGRATION:

$$\rightarrow \underline{x_{n+1}} = \underline{x_n} + \underline{\Delta t} \underline{\dot{x}_{n+1}}$$

(a) EACH TIMESTEP, USE ITERATIVE SOLUTION METHOD DESCRIBED ABOVE FOR NONLINEAR DC ANALYSIS

EXAMPLE: NONLINEAR CAPACITOR

$$\begin{array}{l} I_c \downarrow \frac{1}{T} \begin{array}{c} + \\ - \end{array} V_c \end{array} \rightarrow \underline{Q_c} = \underline{f(V_c)}$$
$$\rightarrow \underline{I_c} = \underline{\dot{Q}_c}$$

$$\rightarrow \underline{Q_{n+1}} = \underline{Q_n} + \Delta t \dot{\underline{Q}}_{n+1}$$

$$\rightarrow \underline{I_{n+1}} = \frac{Q_{n+1}}{\Delta t} - \frac{Q_n}{\Delta t}$$

$$\boxed{I_{n+1} = \frac{\underline{f(V_{n+1})}}{\Delta t} - \frac{Q_n}{\Delta t}}$$

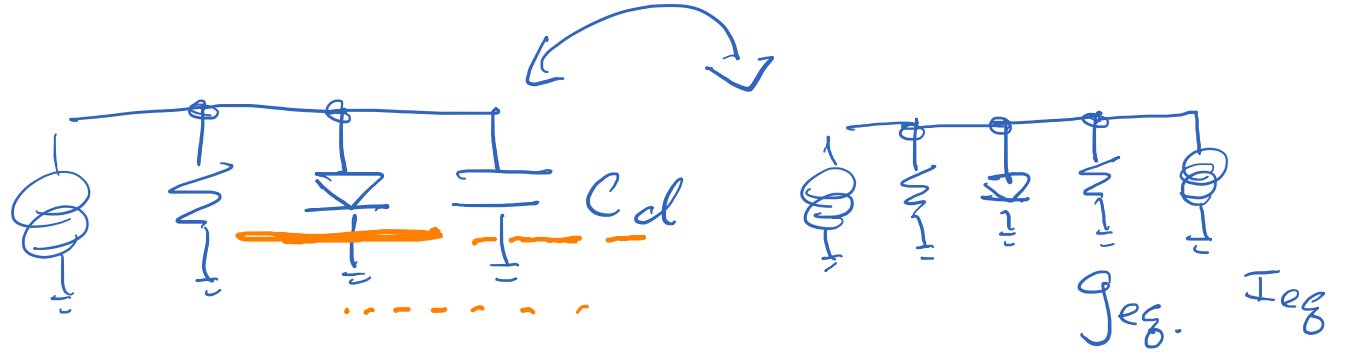
known from
solution
@ t_n

→ APPLY NONLINEAR

→ METHOD FOR

DC CIRCUITS.

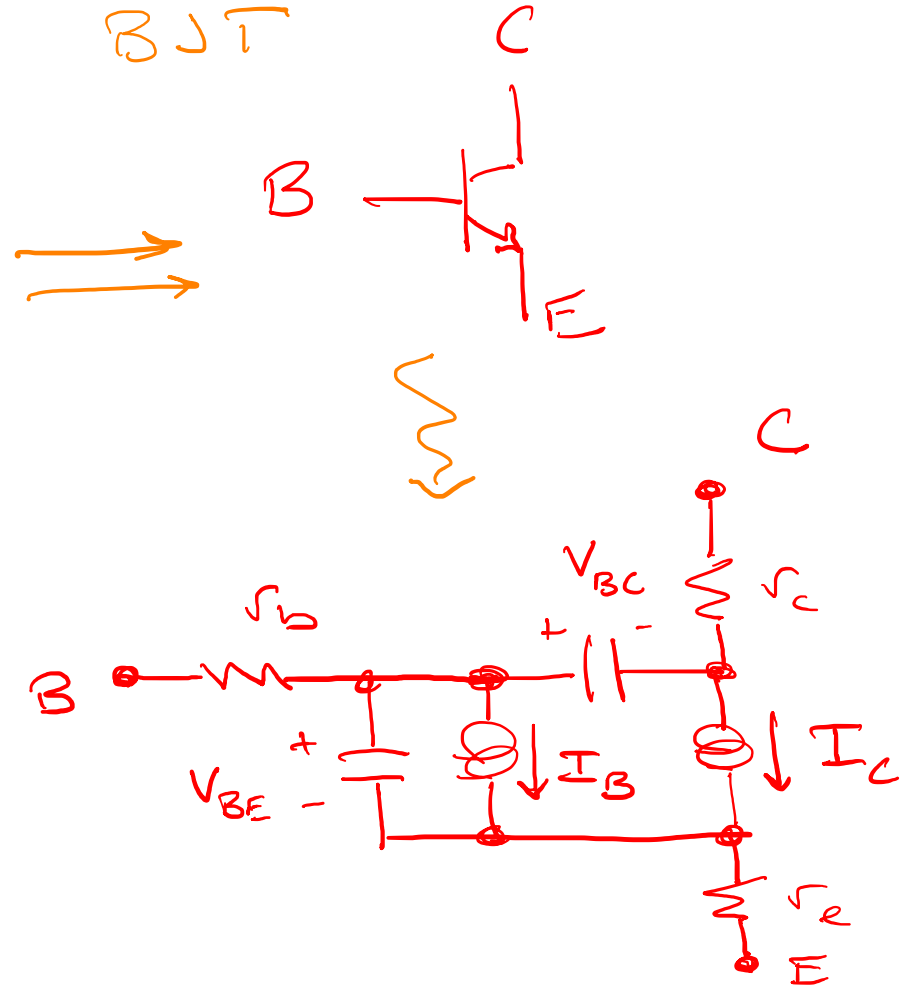
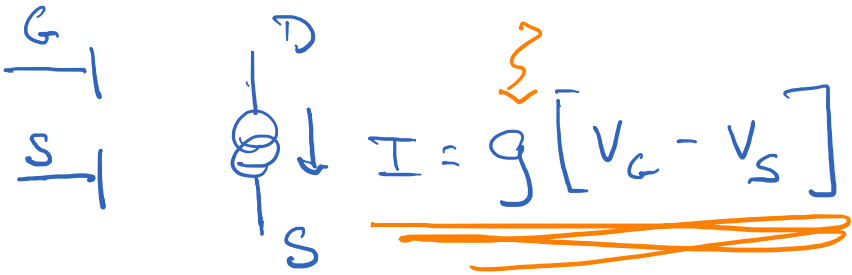
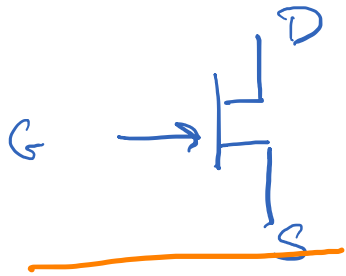
EXAMPLE



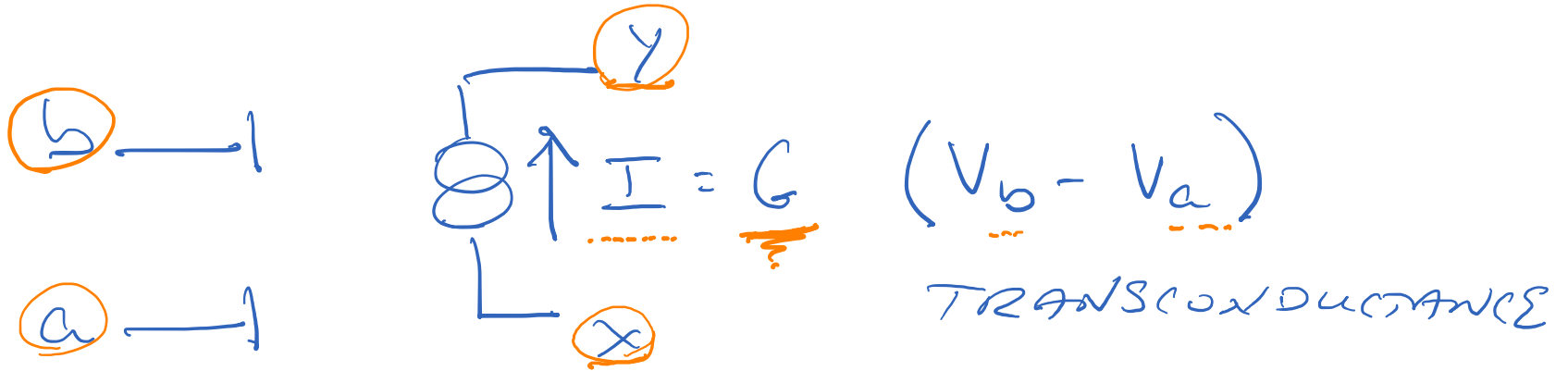
$$\rightarrow C_d = \tau I_s \left[e^{V_c/V_T} - 1 \right]$$

$$\rightarrow \underline{I_{n+1}} = \frac{\tau I_s}{\Delta t} \left[e^{V_{n+1}/V_T} - 1 \right] - \frac{Q_n}{\Delta t}$$

3/4 TERMINAL DEVICES?



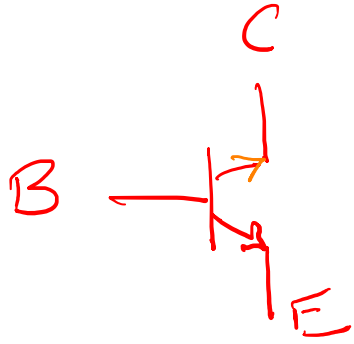
GENERALLY,



G

	a	b	x	y
a	0	0	0	0
b	0	0	0	0
→ x	$+G$	0	$-G$	0
→ y	$-G$	0	$+G$	0

$\left[\begin{array}{c} V_a \\ V_b \\ V_x \\ V_y \end{array} \right]$



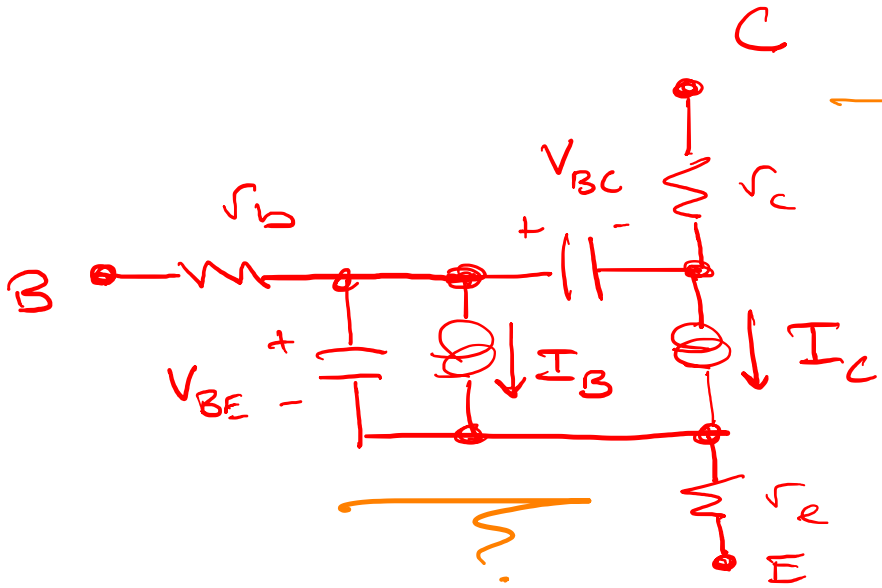
Ebers-Moll

~~Equation~~

$$I_c = I_s \left[e^{V_{BE}/V_T} - e^{V_{BC}/V_T} \right]$$

$$I_B = \frac{I_s}{\beta} \left[e^{V_{BE}/V_T} - 1 \right]$$

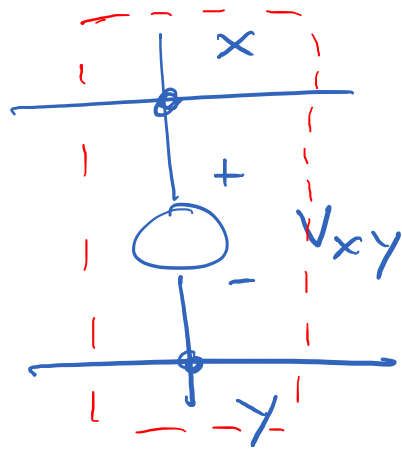
[SIMPLIFIED]



VOLTAGE SOURCES



MODIFIED NODAL ANALYSIS



"SUPER-NODE"

3/4 TERMINAL DEVICES?

