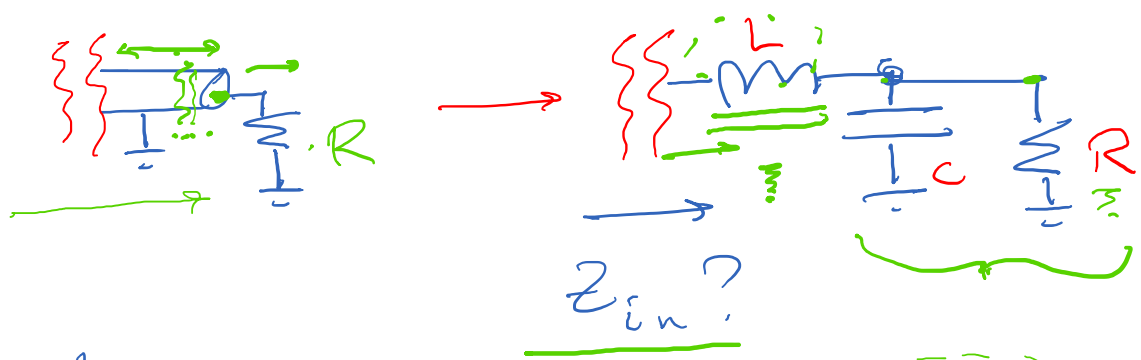


TERMINATED TRANSMISSION LINE

CONSIDER



$$L = L' \Delta x$$

$$C = C' \Delta x$$

LET

$$Z_L = \omega L$$

$$Z_C = 1/\omega C$$

$$Z_{in} = jZ_L - \frac{jZ_C R}{R - jZ_C}$$

$$\frac{1}{1+x} \approx 1-x \quad x \ll 1$$

→ FIND L, C S.T.

$$\underline{R} = Z_{in} = jZ_L - \frac{jZ_C R}{R - jZ_C} \approx jZ_L + R \left[1 - j \frac{R}{Z_C} \right]$$

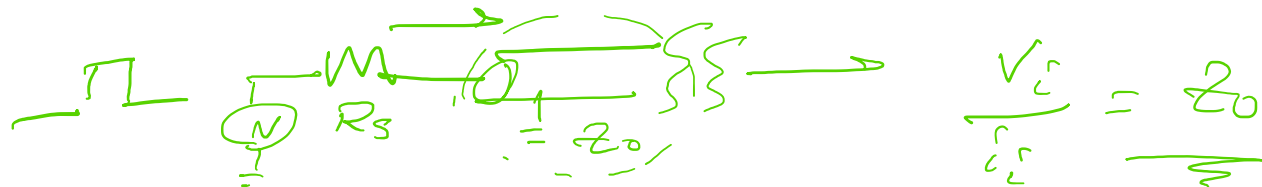
$$\Rightarrow Z_L = \frac{R^2}{Z_C} \quad , \quad \text{OR} \quad R = \sqrt{\frac{L}{C}} = Z_0$$

DEFINE CHARACTERISTIC IMPEDANCE

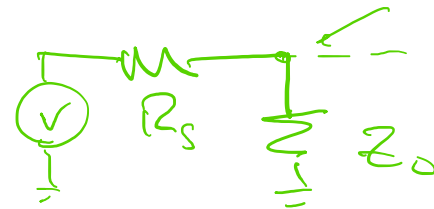
→ $Z_0 = \sqrt{\frac{L'}{C'}}$

IMPEDANCE OF A LOSSLESS LINE TERMINATED

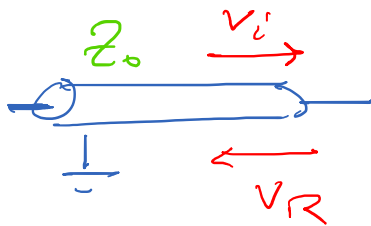
→ $\omega / R = Z_0$ IS PURELY REAL



[ALSO, @ SHORT TIMES, IMPEDANCE LOOKING INTO TRANSMISSION LINE IS PURELY REAL & EQUAL TO Z_0]



THEVENIN EQUIVALENT OF TRANSMISSION LINE



VOLTAGES & CURRENTS ARE SUM OF INCIDENT & REFLECTED WAVES

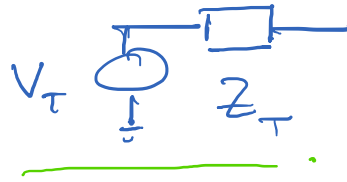
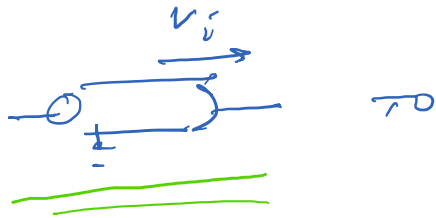
$$\rightarrow V = V_i + V_R$$

$$\rightarrow i = i_i + i_R = \frac{V_i}{Z_0} - \frac{V_R}{Z_0}$$

$$i_i = \frac{V_i}{Z_0}$$

$$i_R = -\frac{V_R}{Z_0}$$

WANT TO MAP



RECALL, $V_T = V_{OC}$

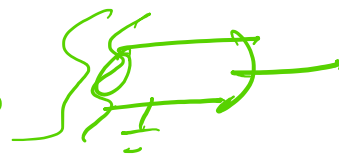
$$Z_T = \frac{V_{OC}}{I_{SC}}$$

V_{oc} ?

IF LINE IS OPEN, $i = 0$

$\rightarrow i_i = -i_R$

$i_i + i_R = 0$

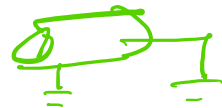


$V_{oc} = V_i + V_R = i_i z_0 - \underbrace{(i_R z_0)}_{+ i_i z_0} = 2 i_i z_0 = 2 V_i$

i_{sc} ?

@ SHORT, $V = 0$

$V_i = -V_R$

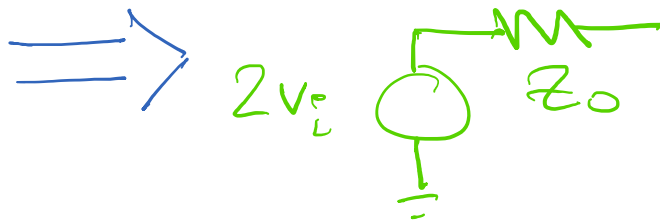
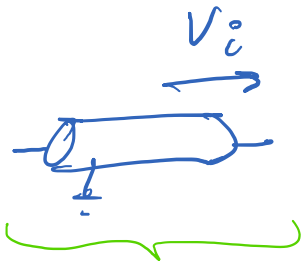
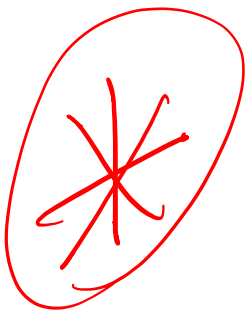


~~V_i~~ $= V_T$

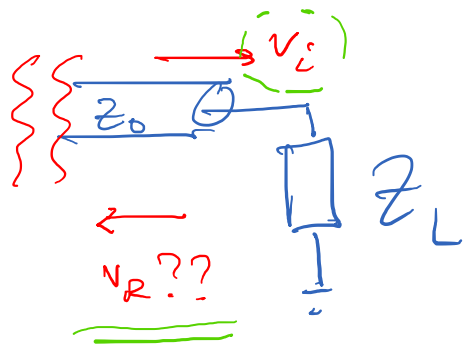
$i = \frac{V_i}{z_0} - \frac{V_R}{z_0} = \frac{2 V_i}{z_0}$

$Z_T = \frac{V_{oc}}{i_{sc}} = \frac{2 V_i}{\left(\frac{2 V_i}{z_0}\right)}$

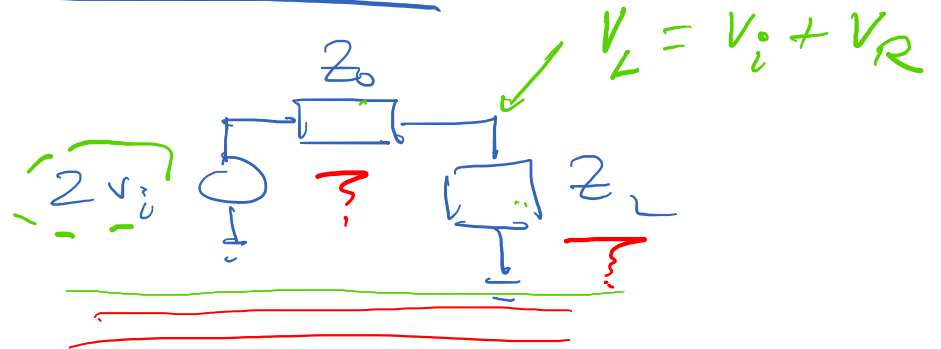
$Z_T = z_0$



REFLECTION FROM MISMATCHED LOAD



\Rightarrow



$$V_L = 2V_i \frac{z_L}{z_L + z_0} = V_i + V_R \dots$$

$$V_R = V_i \left[\frac{2z_L}{z_L + z_0} - \frac{z_L + z_0}{z_L + z_0} \right] = \frac{z_L - z_0}{z_L + z_0} \cdot V_i$$

$$\Gamma = \frac{V_R}{V_i} = \frac{z_L - z_0}{z_L + z_0}$$

CASE 1: $z_L = z_0 \rightarrow \Gamma = 0$

CASE 2: $z_L \rightarrow \infty \rightarrow \Gamma = +1$

CASE 3: $z_L \rightarrow 0 \rightarrow \Gamma = -1$

$\rightarrow \Gamma$

$\leftarrow \Gamma$

IMPEDANCE LOOKING INTO MISMATCHED LINE

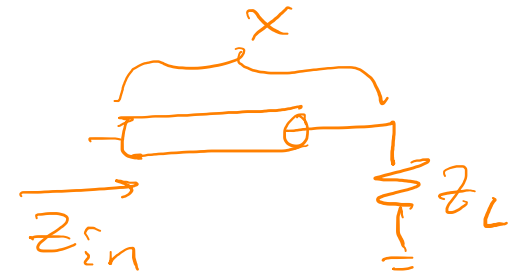
$$V = V_i \left[\underline{e^{\gamma(kx - \omega t)}} + \Gamma \underline{e^{\gamma(-kx - \omega t)}} \right]$$

$$\rightarrow \underline{I} = \frac{V_i}{Z_0} \left[e^{\gamma(kx - \omega t)} - \Gamma e^{\gamma(-kx - \omega t)} \right]$$

$$\underline{Z}_{in} = Z_0 \left[\frac{e^{\gamma kx} + \Gamma e^{-\gamma kx}}{e^{\gamma kx} - \Gamma e^{-\gamma kx}} \right] = Z_0 \left[\frac{1 + \Gamma e^{-2\gamma kx}}{1 - \Gamma e^{-2\gamma kx}} \right]$$

$$\Gamma = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$\boxed{Z_{in} = Z_0 \frac{Z_L + j Z_0 \tan kx}{Z_0 + j Z_L \tan kx}}$$



$$Z_{in} = Z_0 \frac{Z_L + j Z_0 \tan kx}{Z_0 + j Z_L \tan kx}$$

~~case 1~~

CASE 1, $Z_L = Z_0 \implies Z_{in} = Z_0$

CASE 2

~~$kx = \pi/2$~~ $\tan kx \rightarrow \infty$

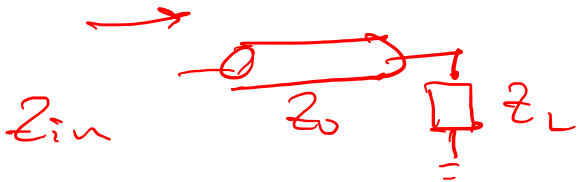
$$Z_{in} = Z_0 \times \frac{j Z_0 \text{ (STAY BIG)}}{j Z_L \text{ (STAY BIG)}} \sim \frac{Z_0^2}{Z_L}$$

$$k = \frac{2\pi}{\lambda}$$

$$\frac{2\pi}{\lambda} \cdot x = \frac{\pi}{2}$$

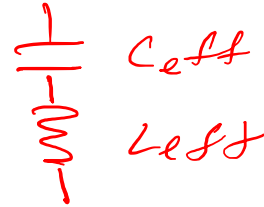
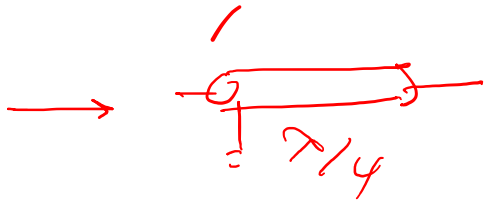
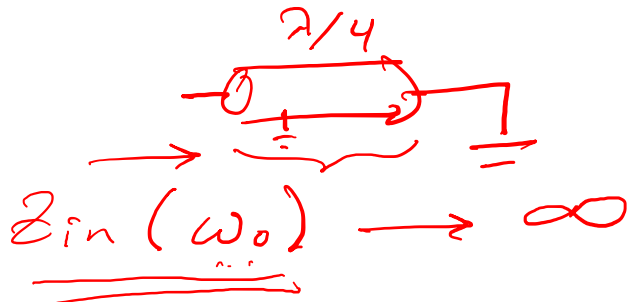
$$x = \lambda/4$$

~~-----~~

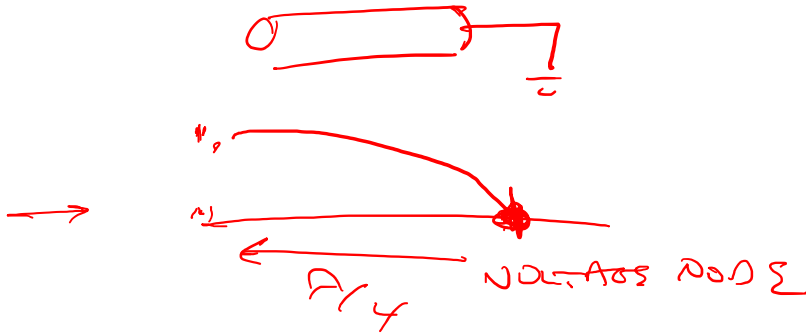


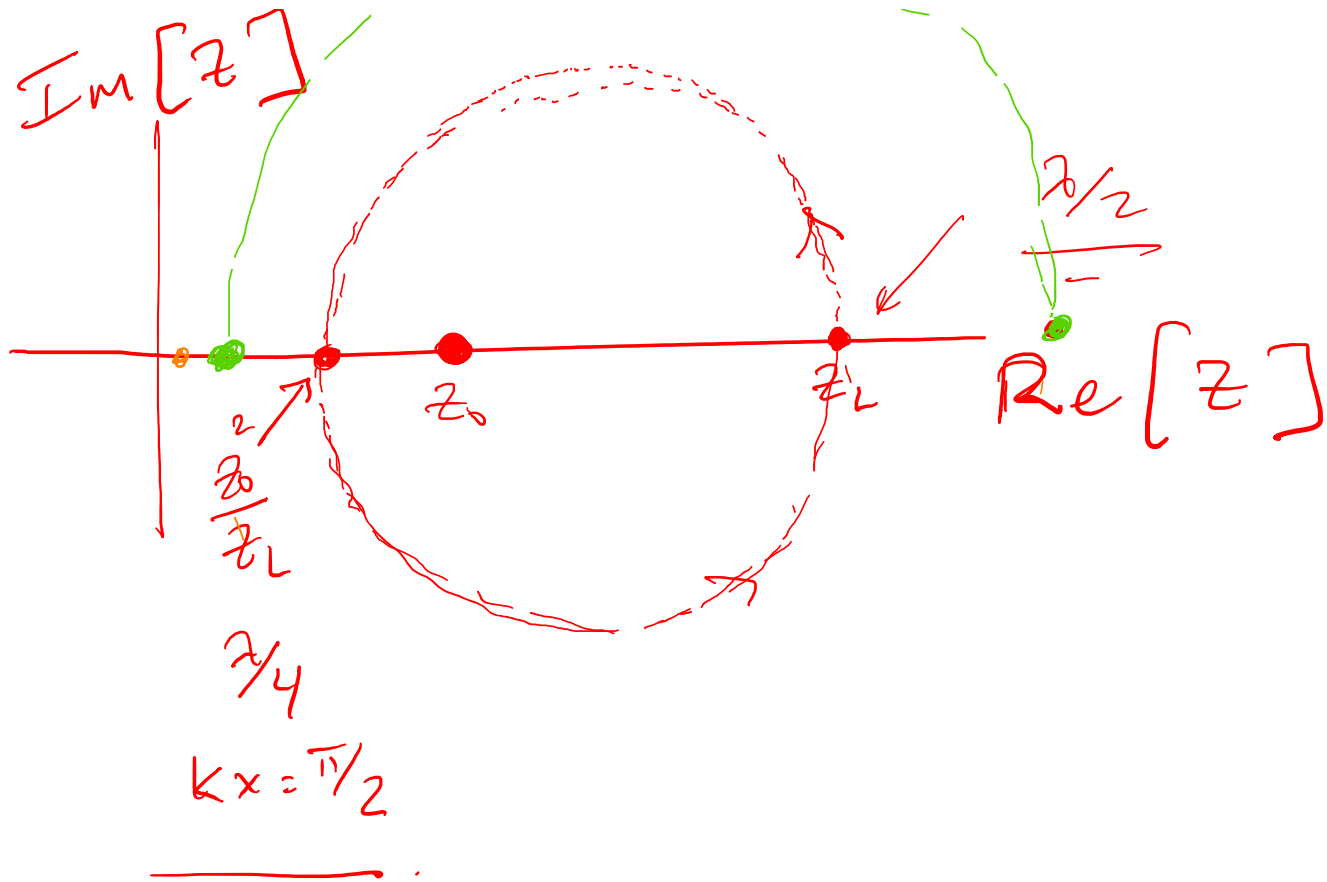
$$Z_L \ll Z_0 \rightarrow @ \lambda/4 \quad Z_{in} = \frac{Z_0^2}{Z_L} \gg Z_0$$

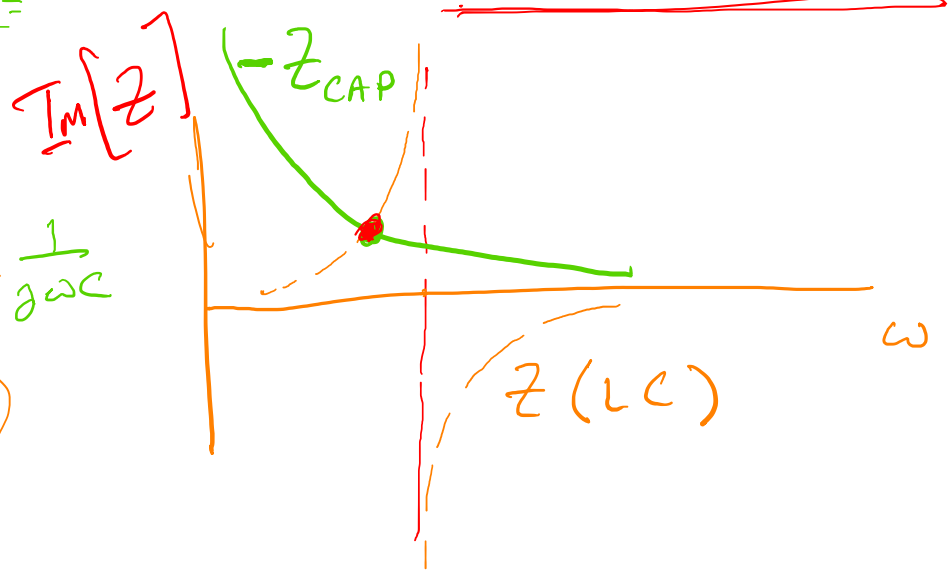
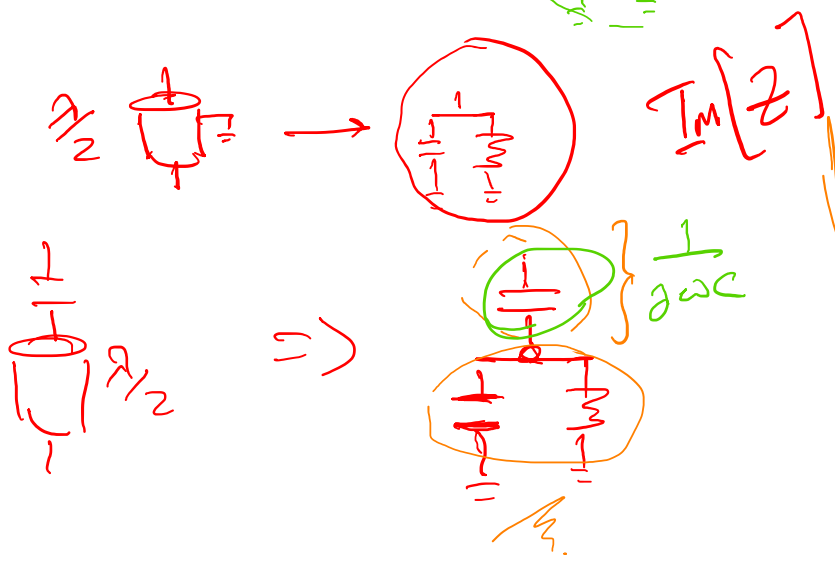
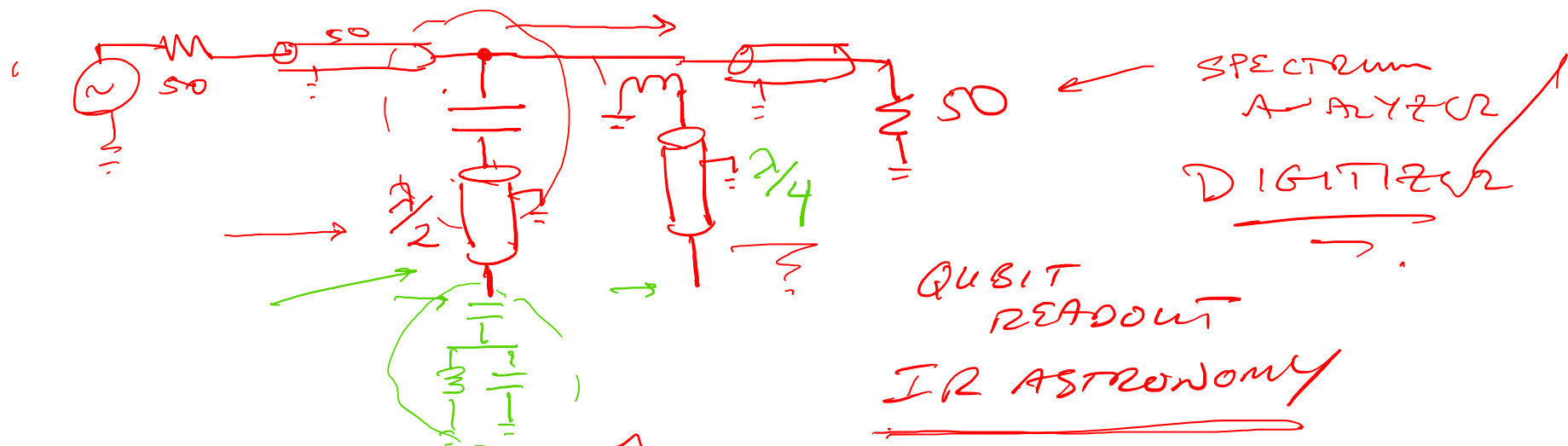
$$Z_L \gg Z_0 \quad @ \lambda/4 \quad Z_{in} = \frac{Z_0^2}{Z_L} \ll Z_0$$

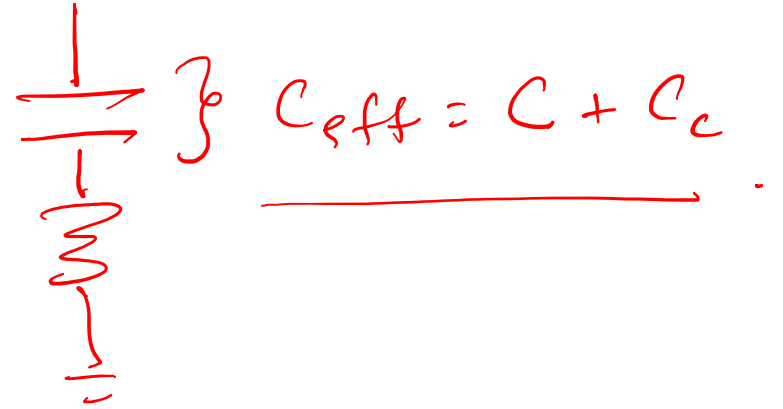
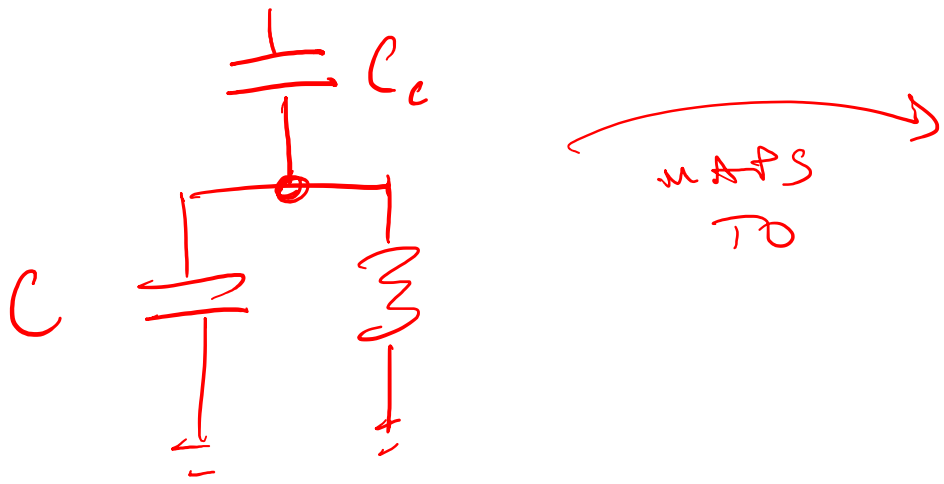


$Z_{left} \approx j \frac{Z_0}{\omega_0}$









BAND-STOP FILTER

