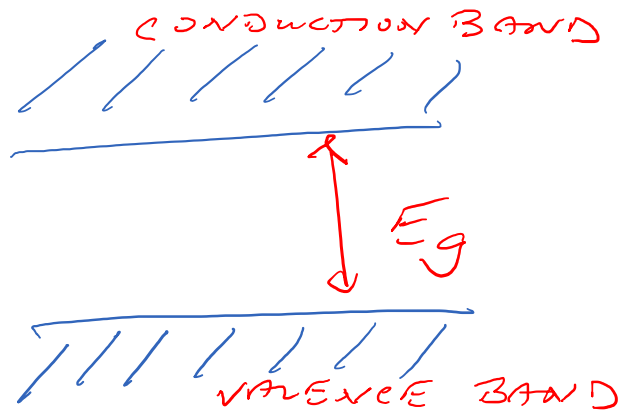


SEMICONDUCTORS

Group IV : Si, Ge

BAND STRUCTURE



$$\text{Si} : E_g = 1.12 \text{ eV}$$

$$\text{Ge} : E_g = 0.66 \text{ eV}$$

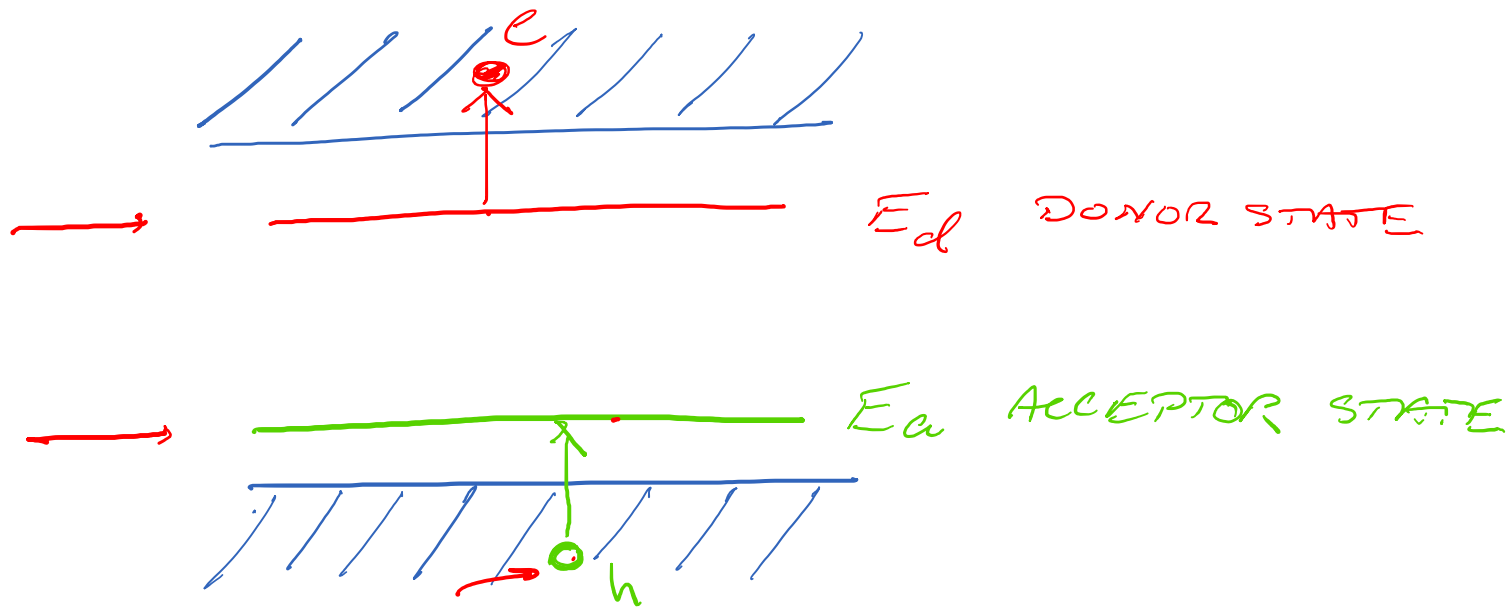
INTRINSIC SEMICONDUCTOR :

$$\text{CARRIER DENSITY} \sim \underline{e^{-E_g/k_B T}}$$

DOPE w/ COL. III or COL. V TO GET CONDUCTION

h
 P-TYPE

e
 N-TYPE

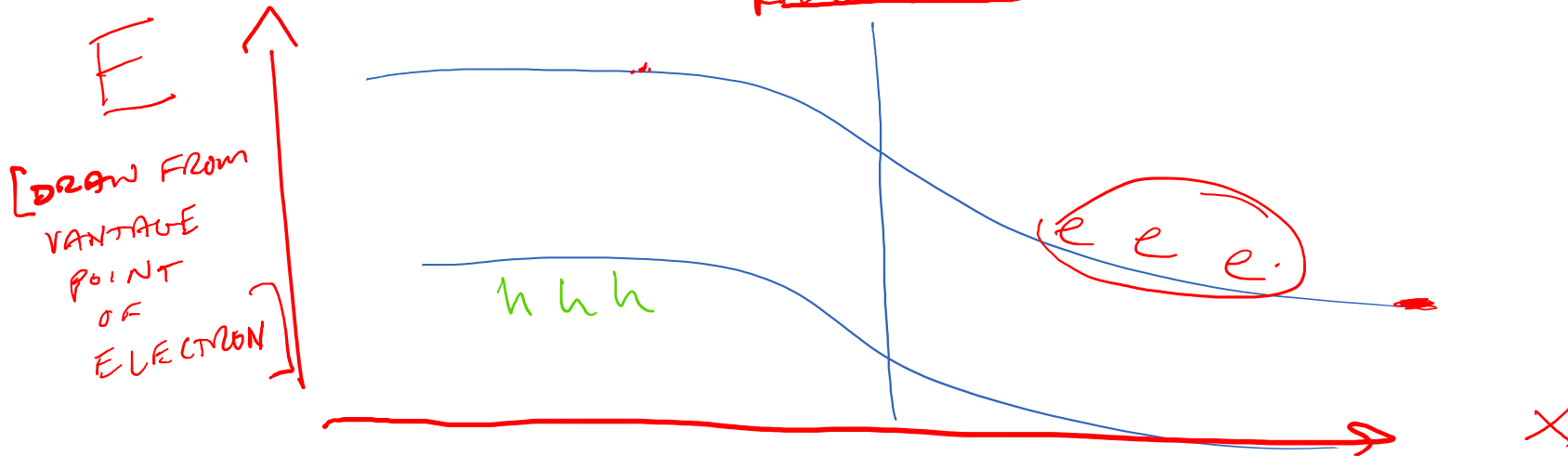
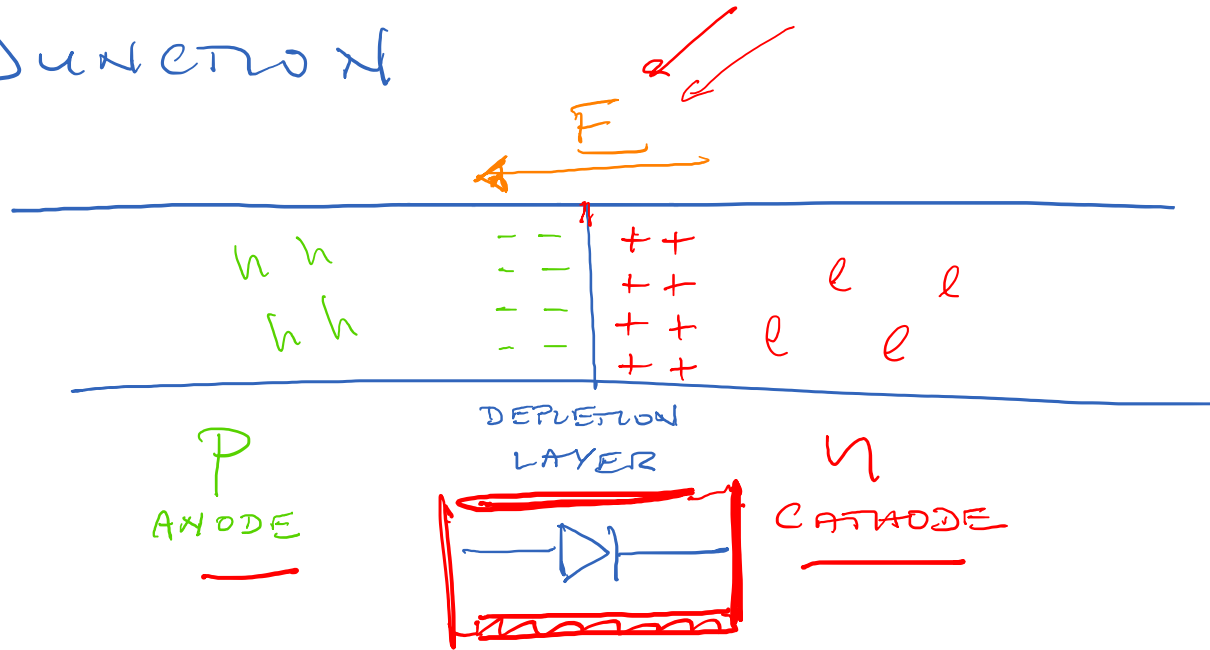


INTRINSIC Si : $\sim 10^{10} \text{ cm}^{-3}$

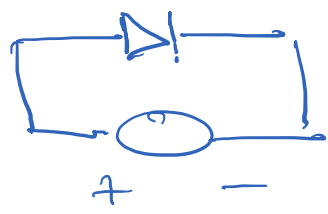
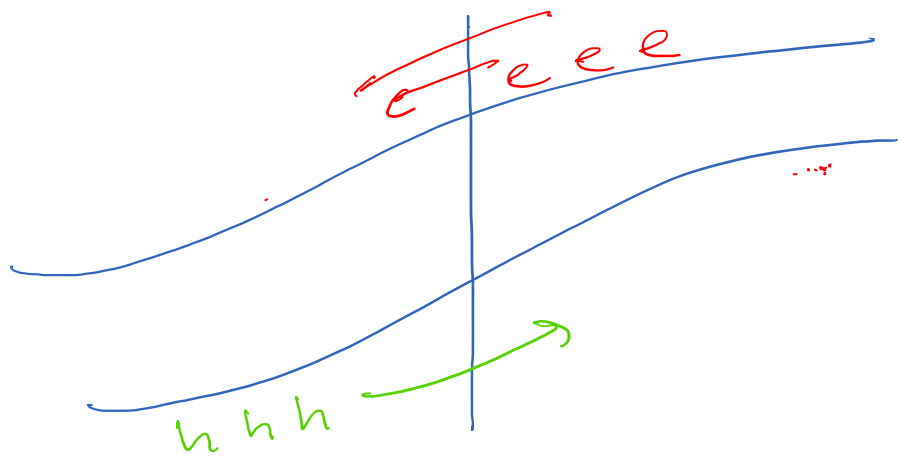
DOPED Si : TO $\sim 10^{18} \text{ cm}^{-3}$

$\left. \begin{array}{l} \triangleleft \\ \right\} \end{array} \right.$

PI Junction

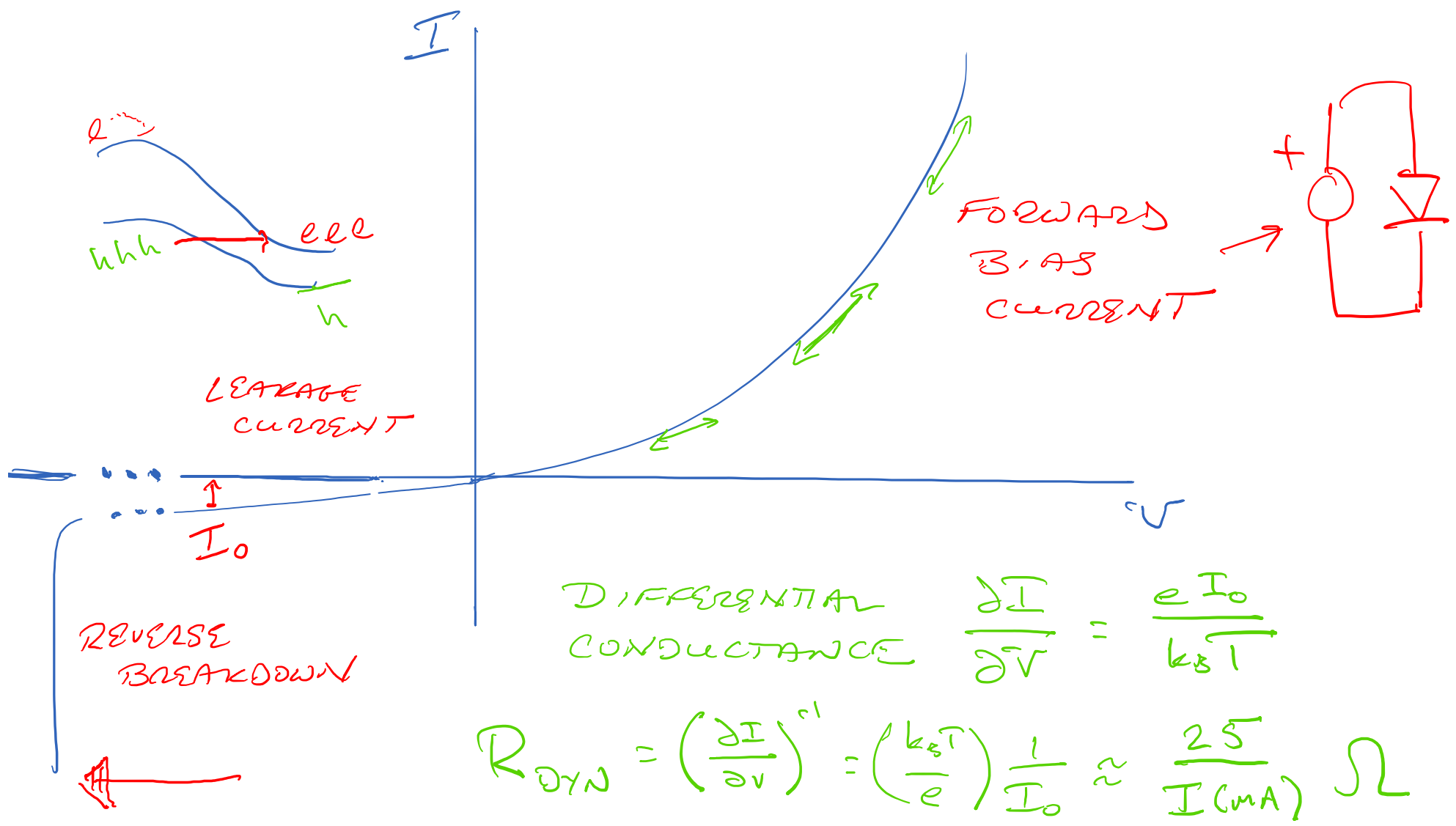


FORWARD
BIAS



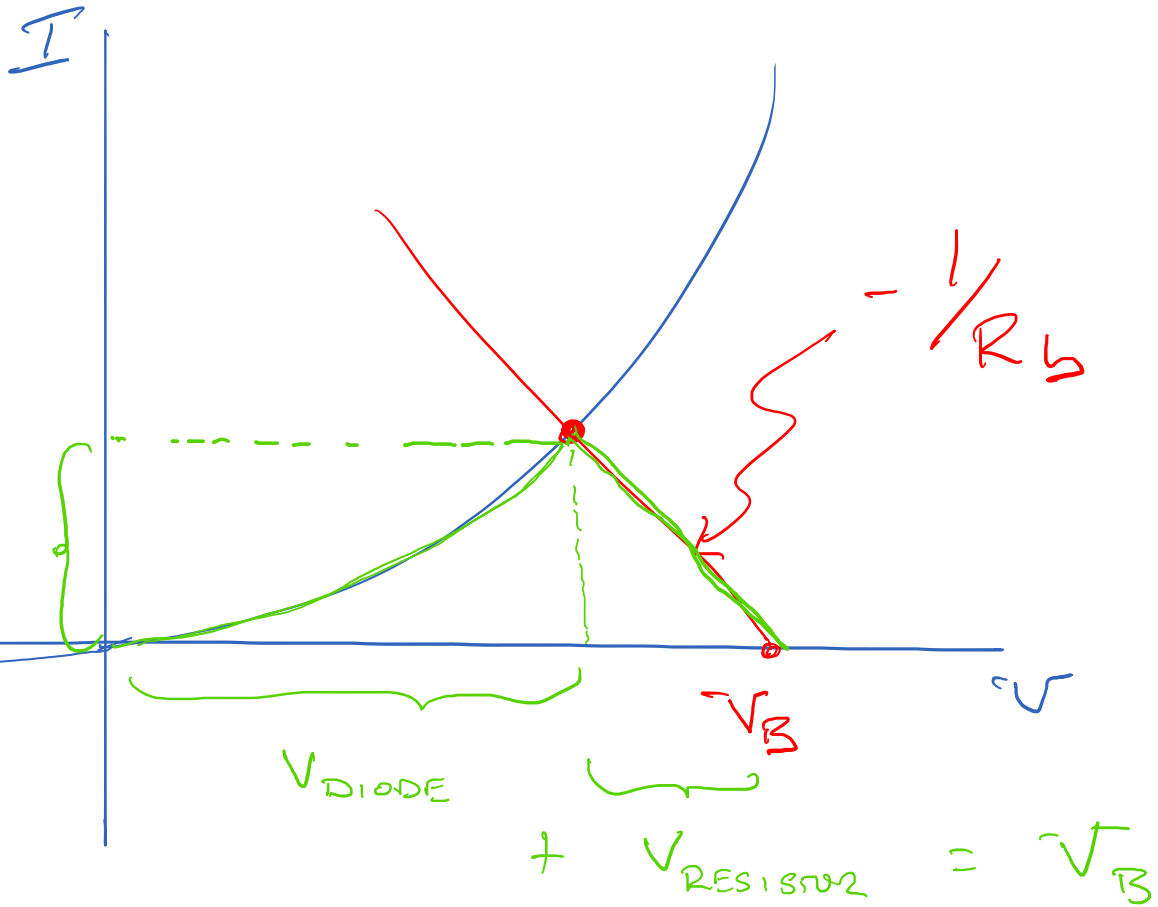
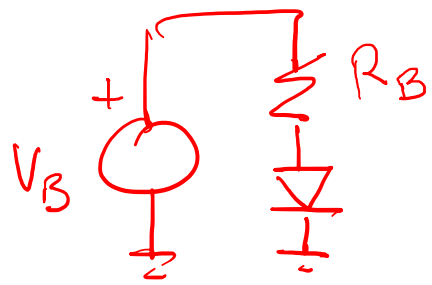
$$I = I_0 \left[e^{\frac{eV}{k_B T}} - 1 \right]$$

$$\frac{\partial I}{\partial V} \approx \frac{e I_0}{k_B T}$$



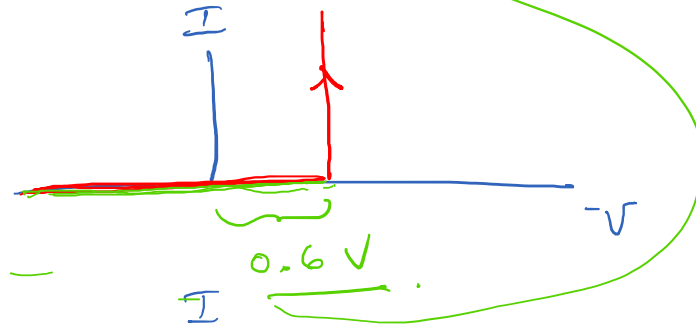
DIFFERENTIAL CONDUCTANCE $\frac{\partial I}{\partial V} = \frac{e I_0}{k_B T}$

$R_{DYN} = \left(\frac{\partial I}{\partial V}\right)^{-1} = \left(\frac{k_B T}{e}\right) \frac{1}{I_0} \approx \frac{25}{I(\text{mA})} \Omega$



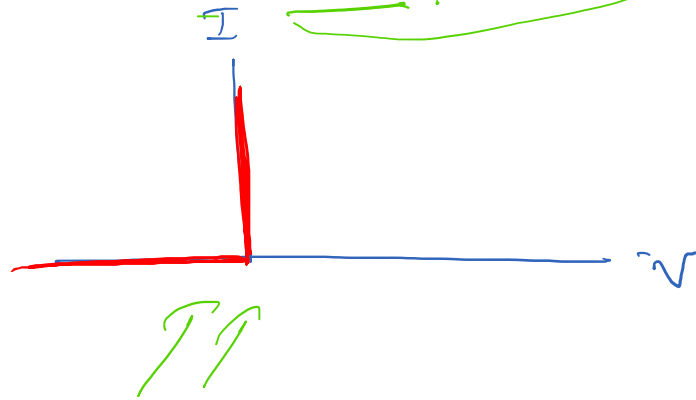
I_0

SIMPLIFIED MODEL

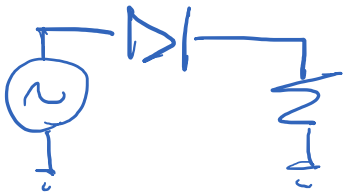


Ge $\sim 0.2 - 0.3V$

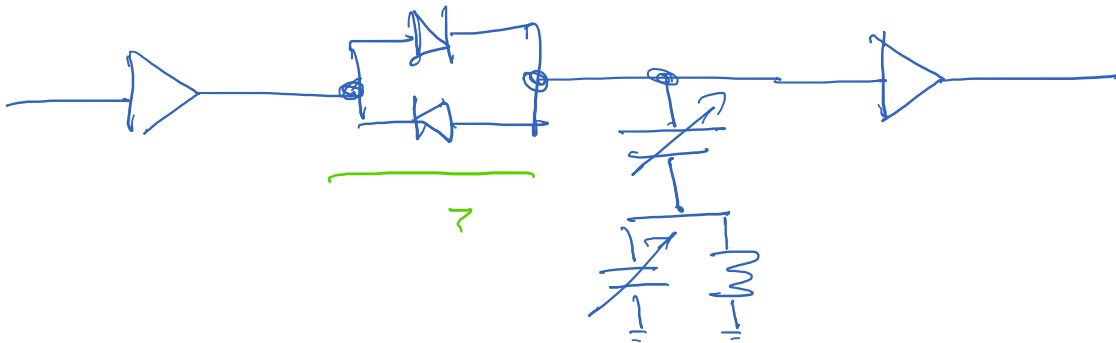
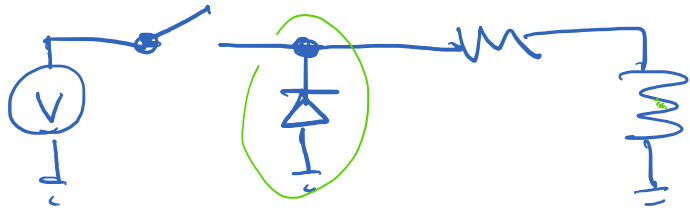
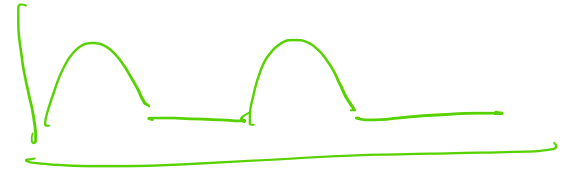
LARGE SIGNALS



SOME APPLICATIONS

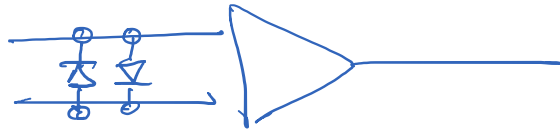


RECTIFIER

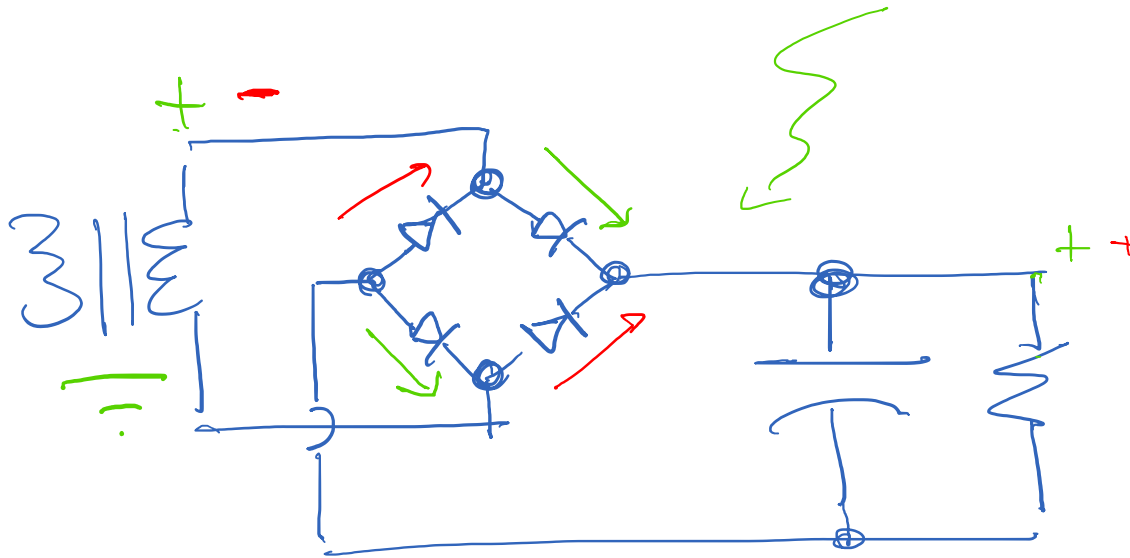


DIODE APPLICATIONS, CONT'D

DIODE
CLAMP

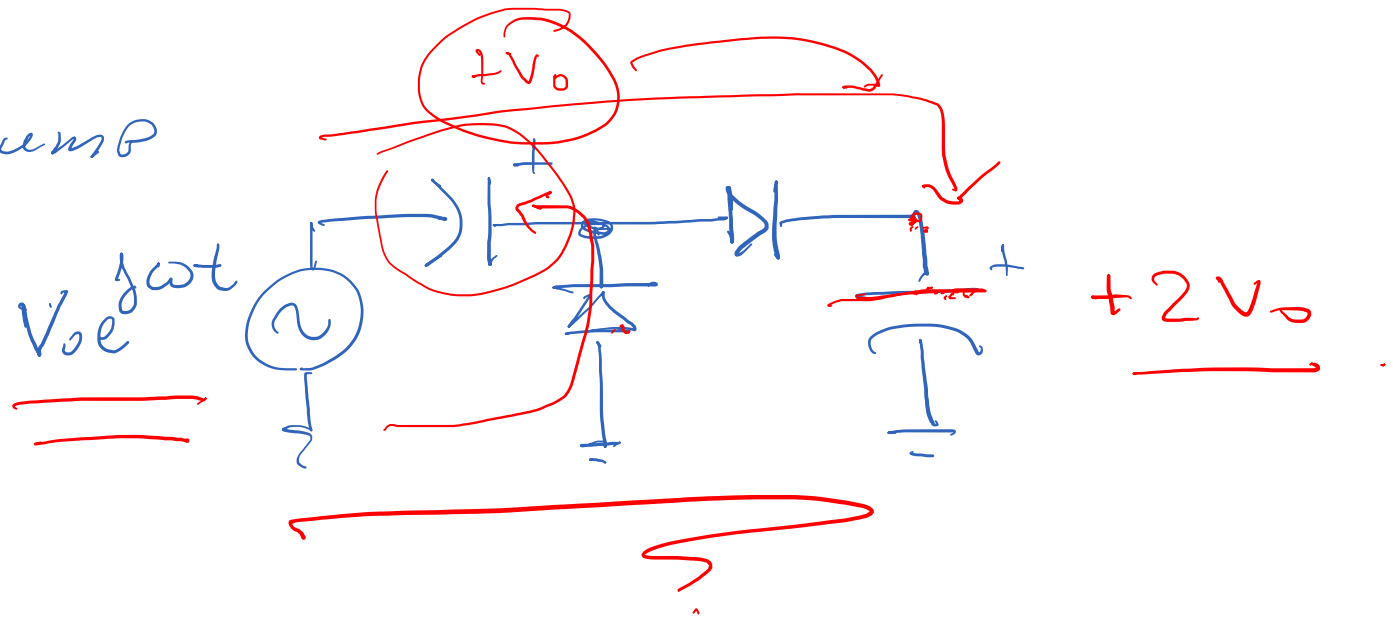


BRIDGE
RECTIFIER

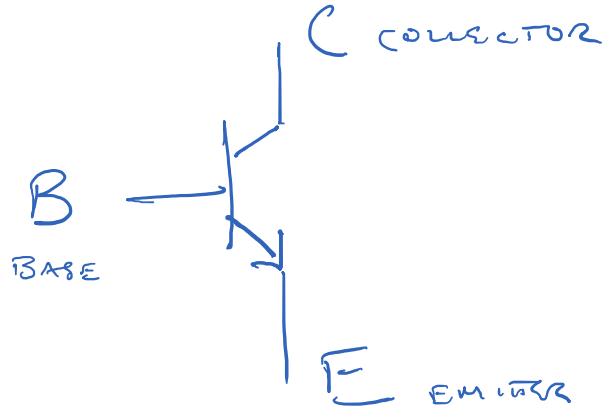
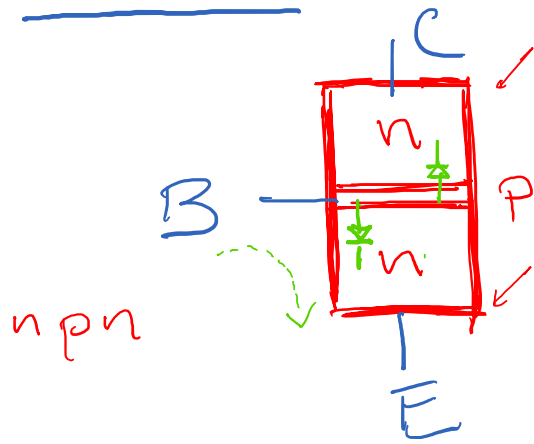


DIODE APPLICATIONS, CONT'D

CHARGE PUMP

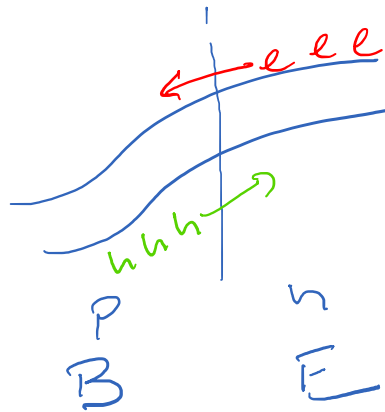


BIPOLAR TRANSISTOR

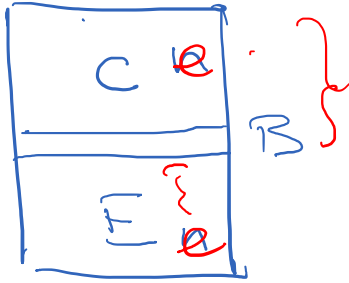


SMALL CURRENT
FROM BASE TO
EMITTER PROVIDES
MUCH LARGER
CURRENT FROM
COLLECTOR TO
EMITTER

FORWARD BIAS



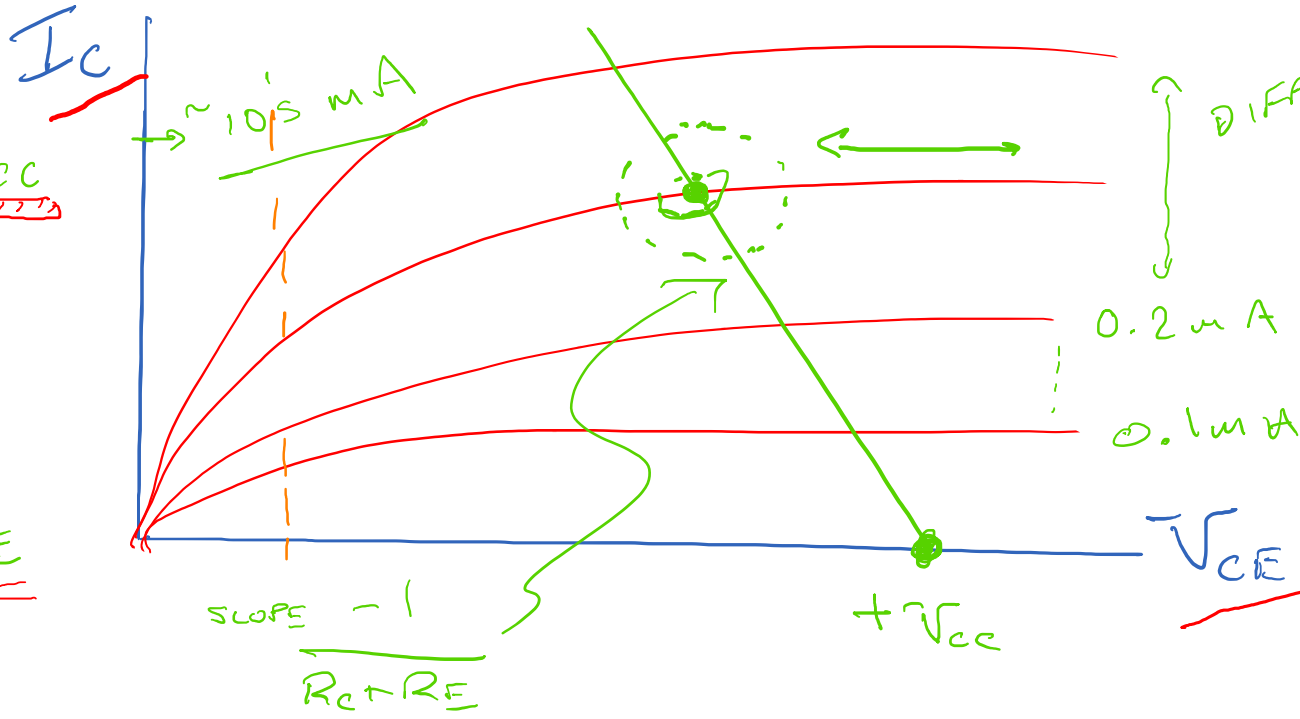
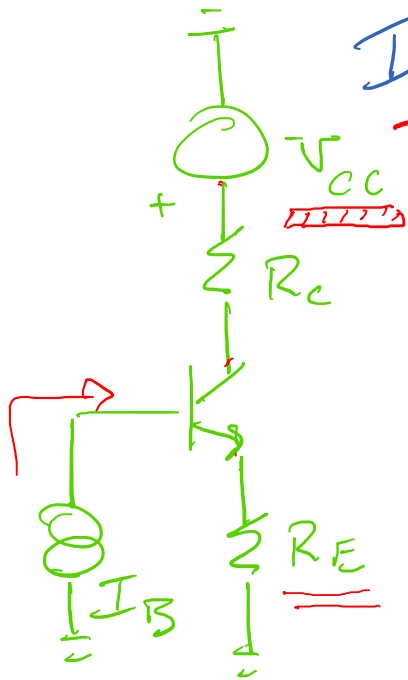
→ CURRENT - CONTROLLED
DEVICE



CB JUNCTION : MAX POWER DISSIPATION

EMITTER : HEAVILY DOPED

BASE : LIGHTLY DOPED
(MINIMIZE RECOMBINATION)



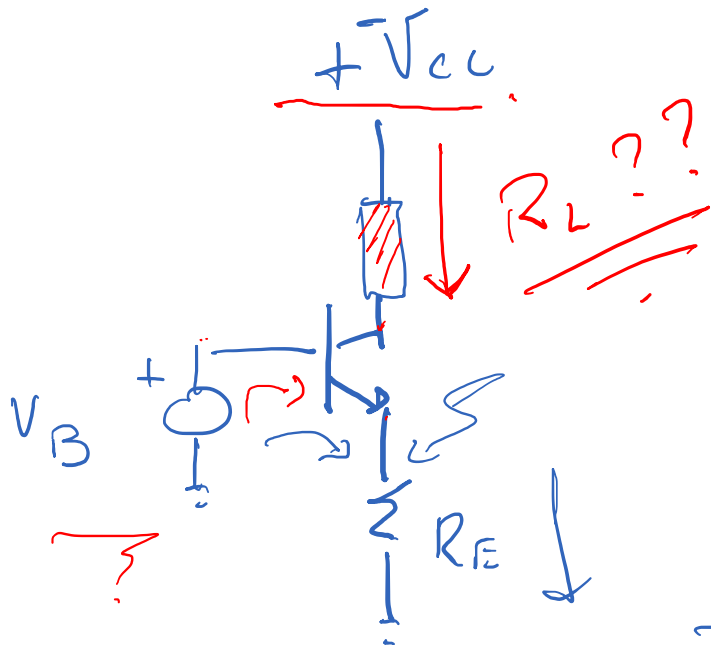
$$\Delta I_C = \beta \Delta I_B$$

$$\beta \sim 100$$

TRANSISTOR CURRENT SOURCE

Current Source

For $V_C \geq 0.2V + V_E$



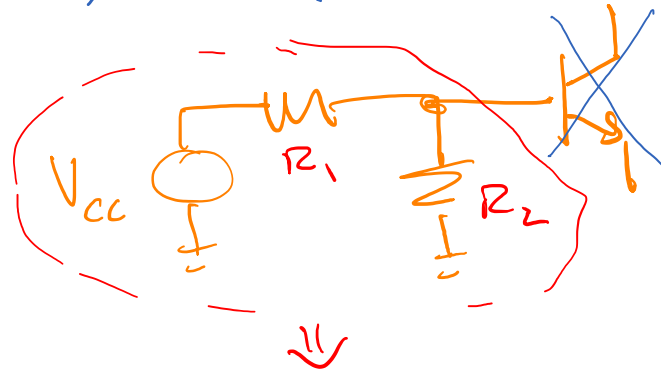
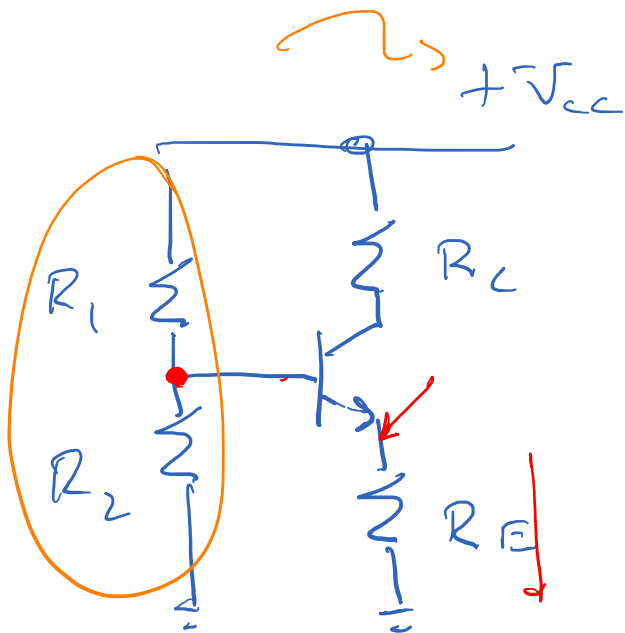
$$V_E \approx V_B - 0.6V$$

$$I_E = \frac{V_B - 0.6V}{R_E}$$

$$I_E = I_B + I_C \approx I_C$$

$\beta \gg 1$

TRANSISTOR BIASING, CONT'D



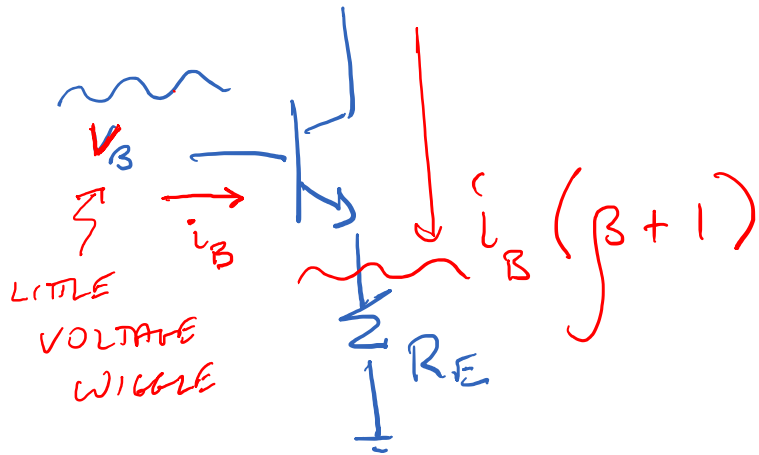
$$V_T = \frac{R_2}{R_1 + R_2} V_{CC}$$

$$R_T = R_1 \parallel R_2 = \frac{R_1 R_2}{R_1 + R_2}$$

NEED

$R_1 \parallel R_2 \ll (\beta + 1) R_E$ ← STIFF VOLTAGE BIAS.

INPUT IMPEDANCE OF BJT?

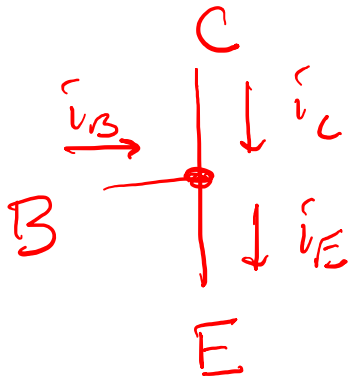


$$R_{in} = \frac{v_{in}}{i_{in}} = \frac{v_B}{i_B}$$

LITTLE WIGGLE @ BASE

IDENTICAL
LITTLE WIGGLE @ EMITTER

$$v_{in} = v_B = v_E = i_E R_E = \underline{\underline{i_B (\beta + 1) R_E}}$$



$$R_{in} = \frac{v_{in}}{i_{in}} = \frac{v_B}{i_B} = (\beta + 1) R_E$$

$$\begin{aligned} i_E &= i_C + i_B \\ &= \beta i_B + i_B = (\beta + 1) i_B \end{aligned}$$

$$R_{in} = (\beta + 1) R_E$$

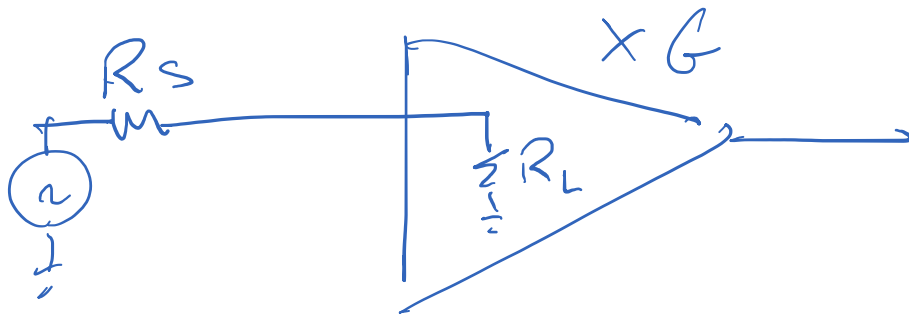


$$V_L = \frac{R_L}{R_s + R_L} \cdot V_s$$

FOR $R_L \gg R_s$

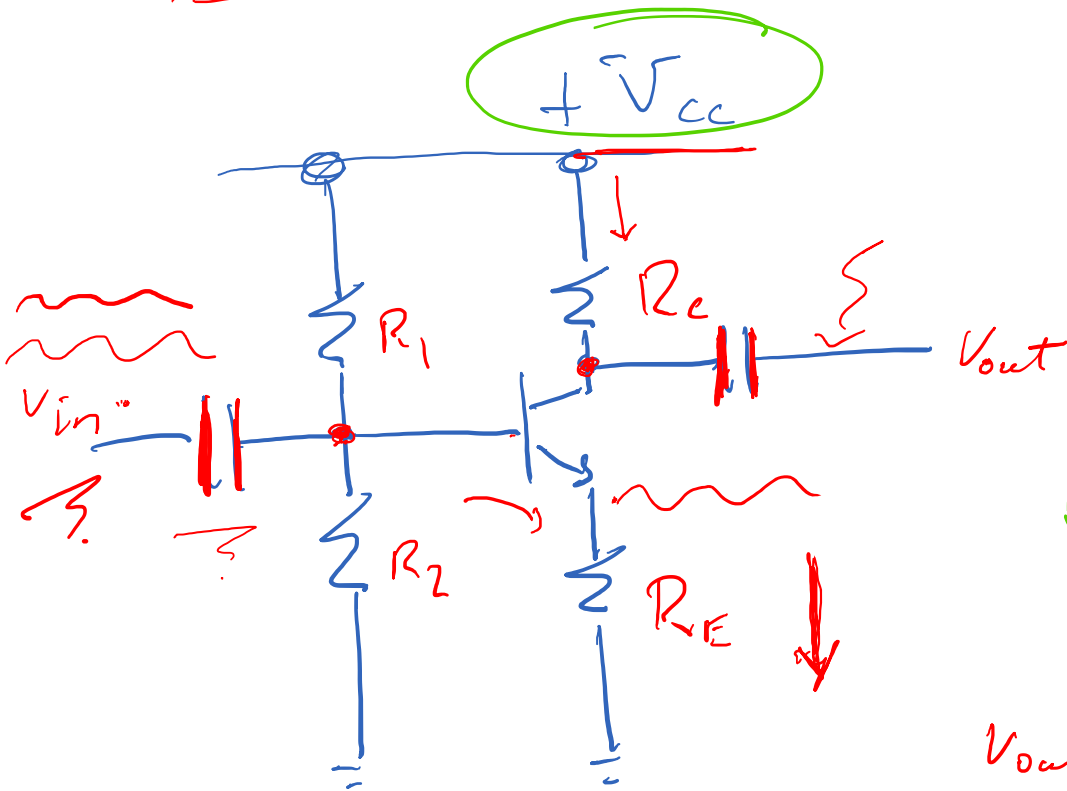
$$V_L \approx V_s$$

* DESIRED,
IF AMPLIFIER
VOLTAGES



"STIFF SOURCE"

COMMON EMITTER AMPLIFIER



$$V_E = V_{in}$$

$$I_E = \frac{V_E}{R_E} = \frac{V_{in}}{R_E}$$

$$I_E \approx I_C \quad [\beta \gg 1]$$

$$I_C \approx \frac{V_{in}}{R_E}$$

$$V_{out} = V_C = -I_C R_C = -\frac{R_C}{R_E} V_{in}$$

$$V_C = V_{CC} - I_C R_C \Rightarrow \left| \frac{V_o}{V_i} = -\frac{R_C}{R_E} \right|$$